

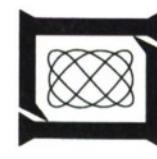
Correlation of Nonlinear Distortion in Digital Phased Arrays: Measurement and Mitigation

D.J. Rabideau
L.C. Howard

Presented at the 2002 IEEE MTT-S International Microwave Symposium, 4 June 2002

Issued 1 November 2004

Lincoln Laboratory
MASSACHUSETTS INSTITUTE OF TECHNOLOGY
LEXINGTON, MASSACHUSETTS



Prepared for the Department of the Air Force under Contract F19628-00-C-0002.

Approved for public release; distribution is unlimited.

101-1037

ADA427551

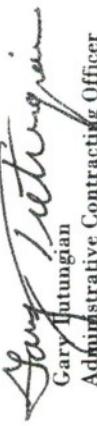
This report is based on studies performed at Lincoln Laboratory, a center for research operated by Massachusetts Institute of Technology. This research effort was sponsored under the auspices of the Lincoln Laboratory Advanced Concepts Program. The Advanced Concepts Program is supported principally by the Department of the Air Force under Contract F19628-00-C-0002.

This report may be reproduced to satisfy needs of U.S. Government agencies.

The ESC Public Affairs Office has reviewed this report, and it is releasable to the National Technical Information Service, where it will be available to the general public, including foreign nationals.

This technical report has been reviewed and is approved for publication.

FOR THE COMMANDER



Gary J. Iltangian
Administrative Contracting Officer
Plans and Programs Directorate
Contracted Support Management

Non-Lincoln Recipients
PLEASE DO NOT RETURN
Permission is given to destroy this document
when it is no longer needed.

Correlation of Nonlinear Distortion in Digital Phased Arrays: Measurement and Mitigation

Daniel J. Rabideau
L. Cole Howard

Lincoln Laboratory, Massachusetts Institute of Technology

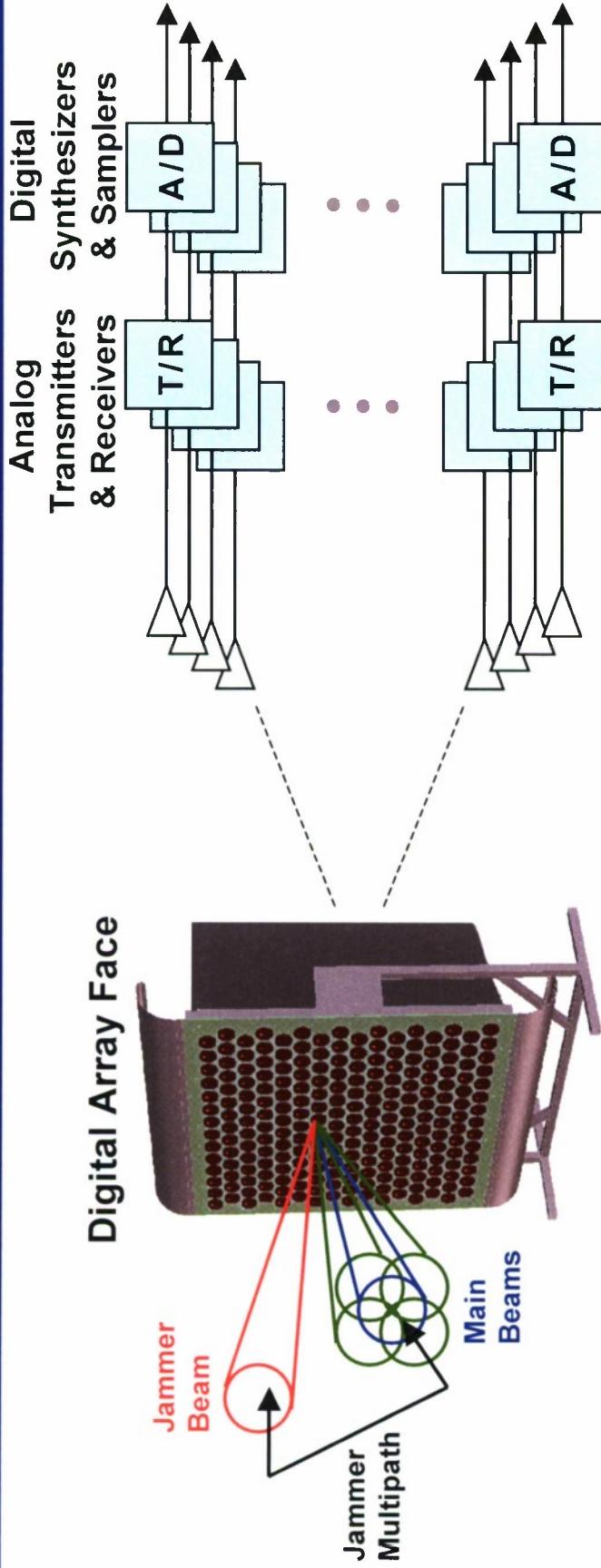
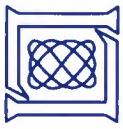
Presented at the 2002 IEEE MTT-S International Microwave Symposium, 4 June 2002

This work was sponsored under Air Force Contract F19628-00-C-0002.
Opinions, interpretations, conclusions and recommendations are those of the author
and are not necessarily endorsed by the Department of Defense.

50
years



Digital Arrays & Their Benefits



- Digital sampling & filtering behind each traditional analog receiver
 - Digital Synthesis (DDS) in front of each traditional analog transmitter
- Enables:
 - Multiple simultaneous beams; Adaptive beams
 - Flexible scan patterns; Faster area search
 - **Improved dynamic range & linearity**
 - Potential for modular, open-systems design

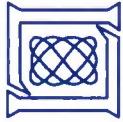
*The subject of
this presentation*



Outline

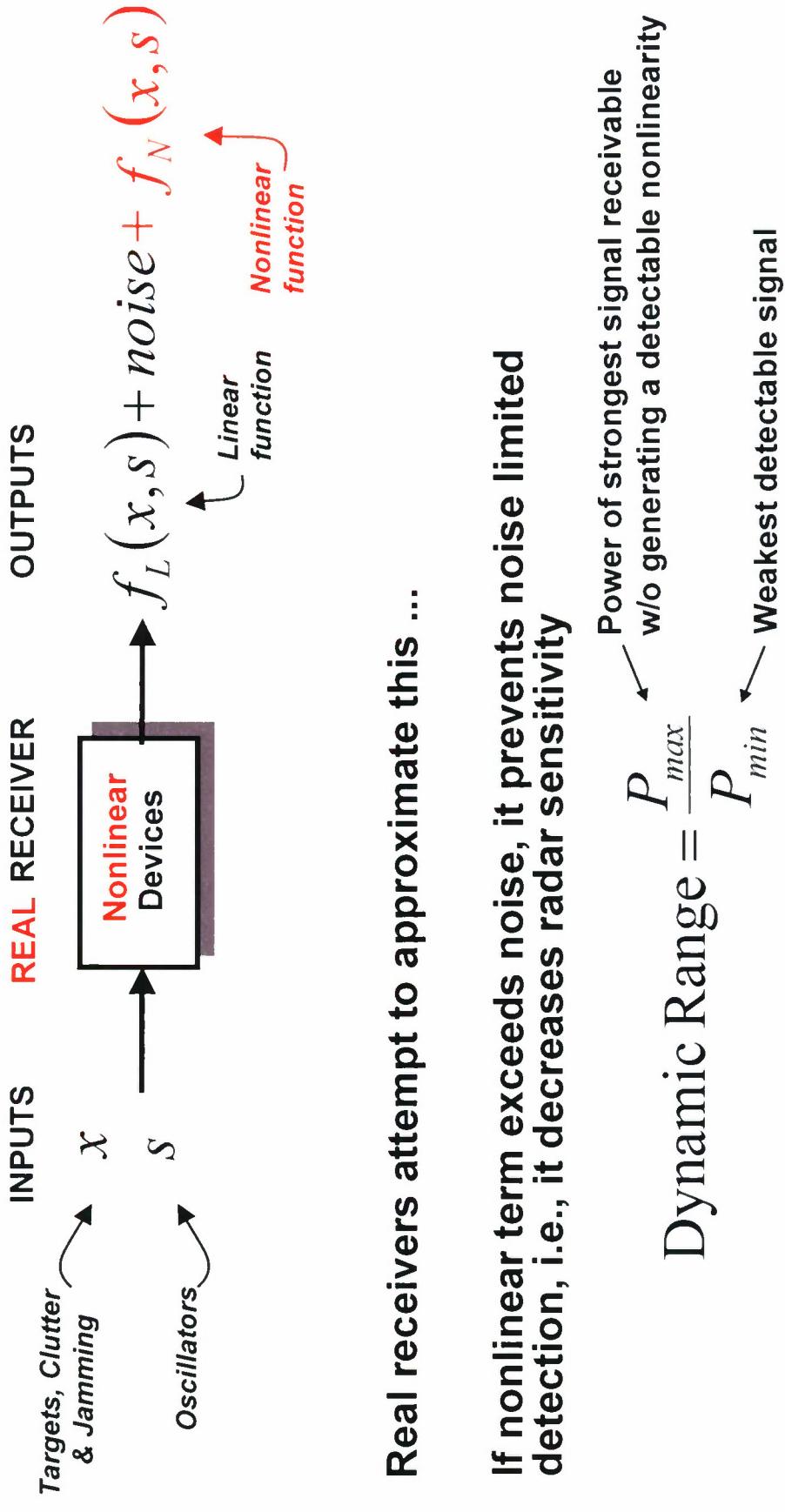
- **Introduction**
- **Importance of linearity and dynamic range**
- **Method for improving dynamic range**
- **Experimental data collection**
- **Summary and future work**





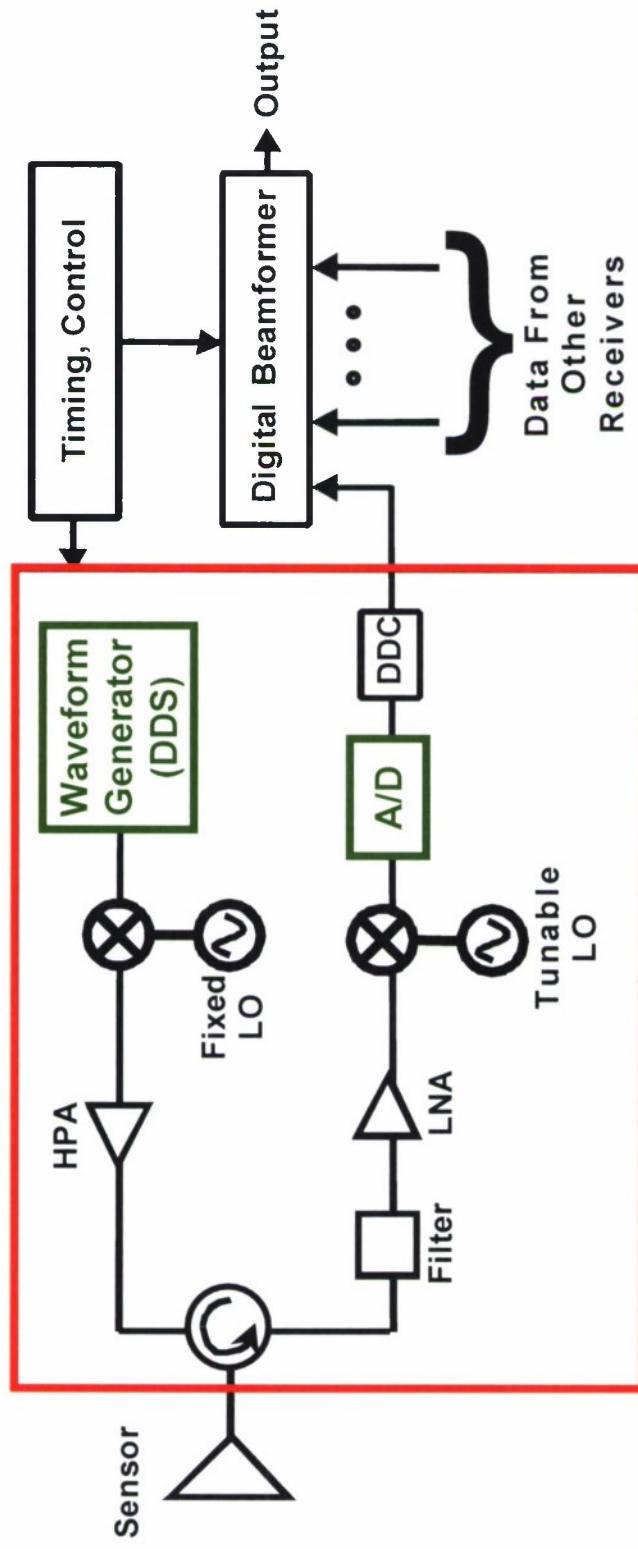
Linearity and Dynamic Range

- An idealized receiver performs linear operations

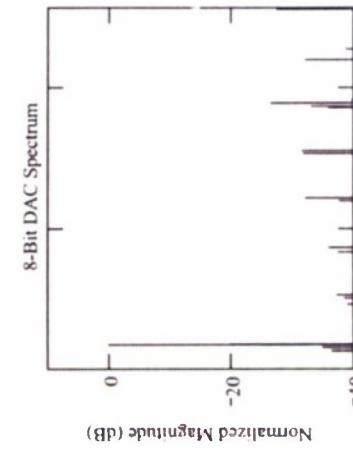
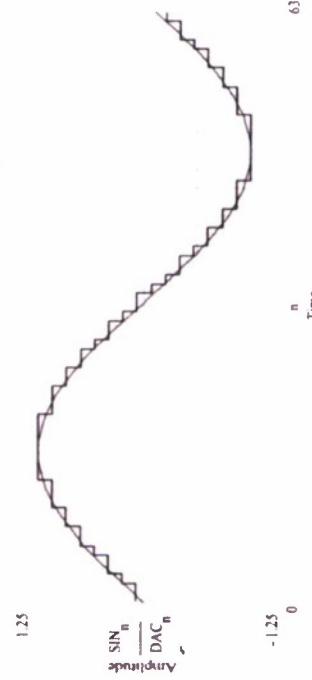




Some Types of Nonlinearity in a Typical Digital T/R Module

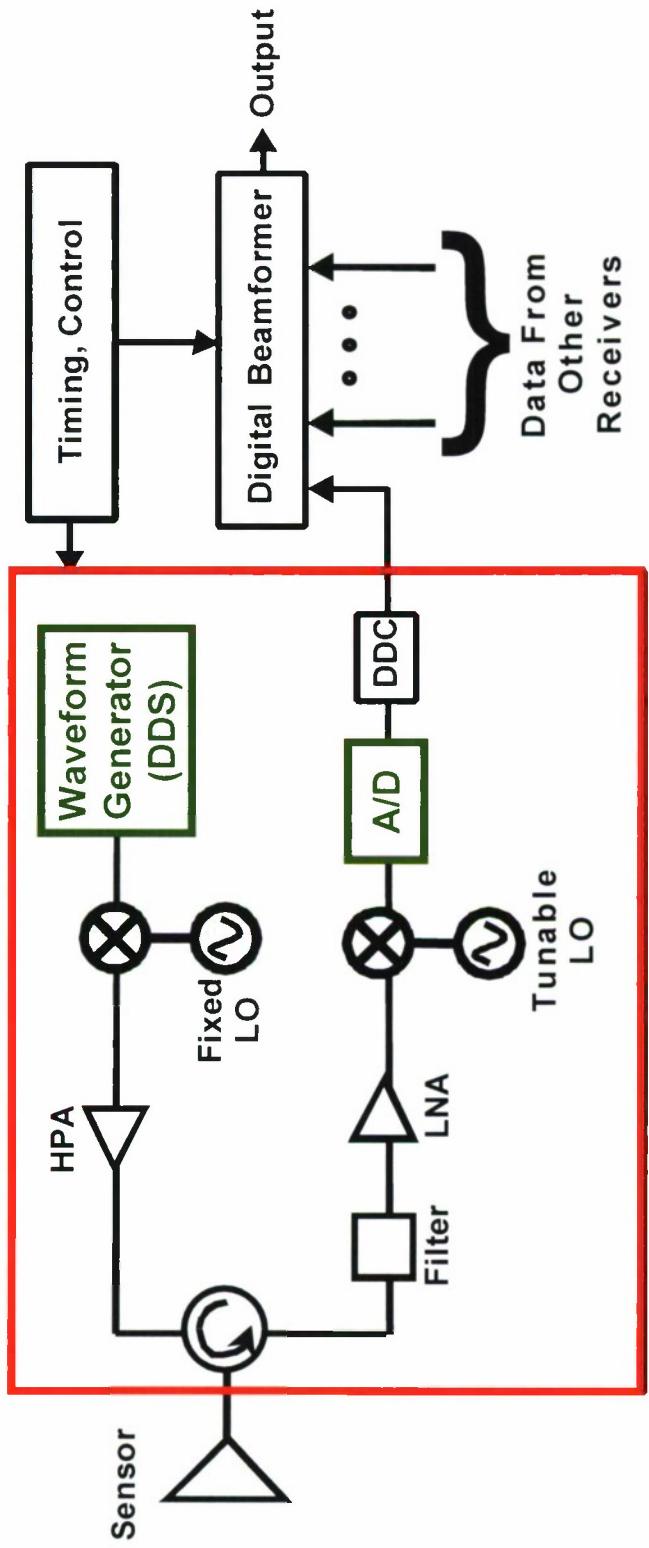


- **Quantization (A/DSs, DACs)**

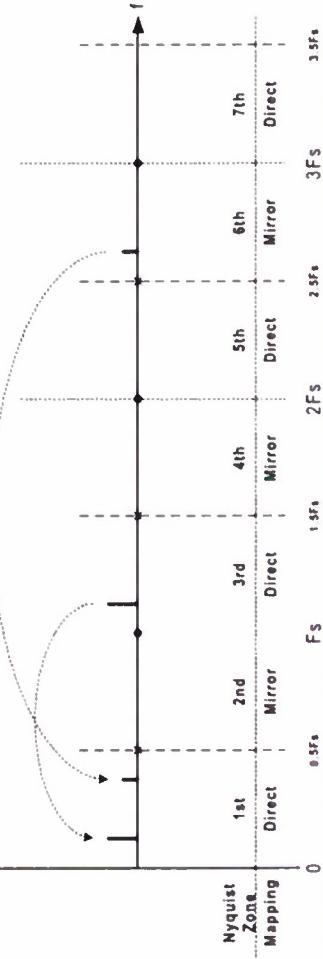




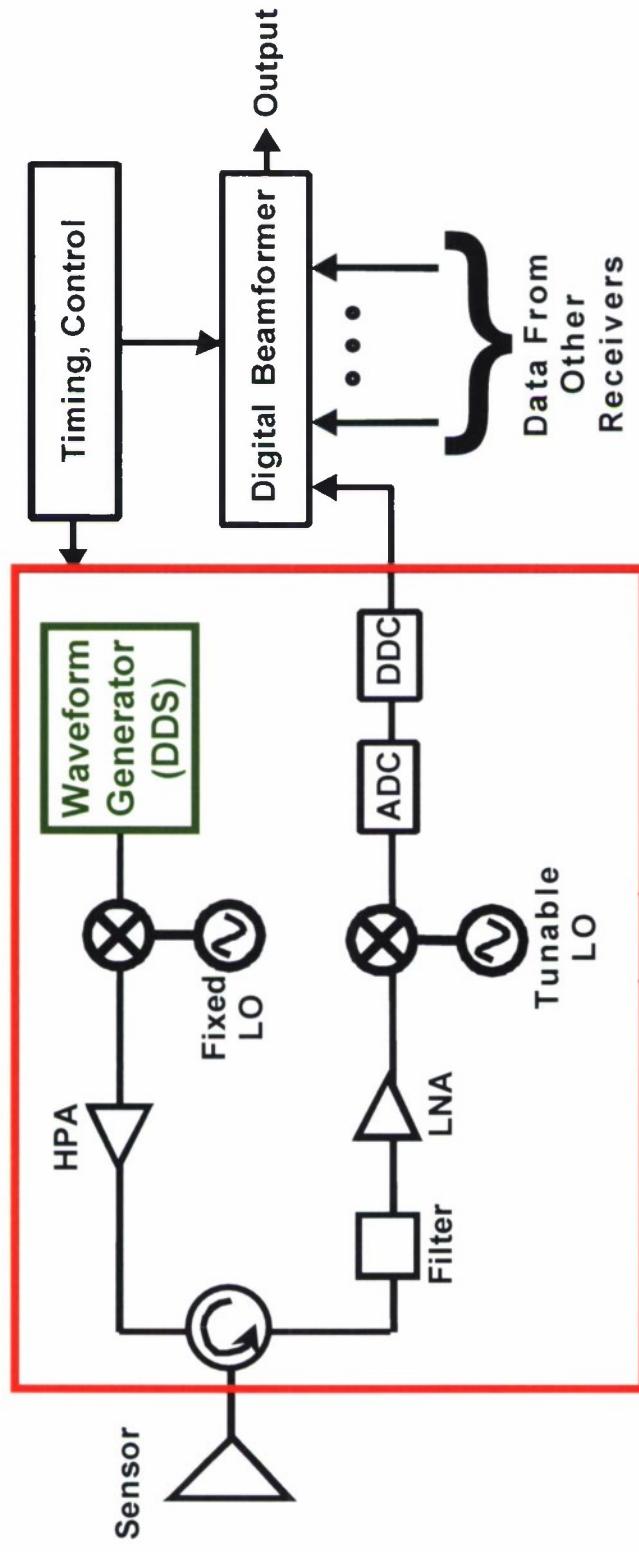
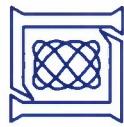
Some Types of Nonlinearity in a Typical Digital T/R Module



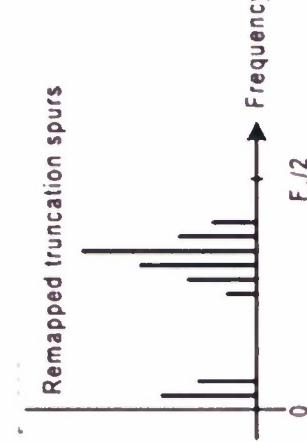
- Harmonic distortions, differential and integral nonlinearities



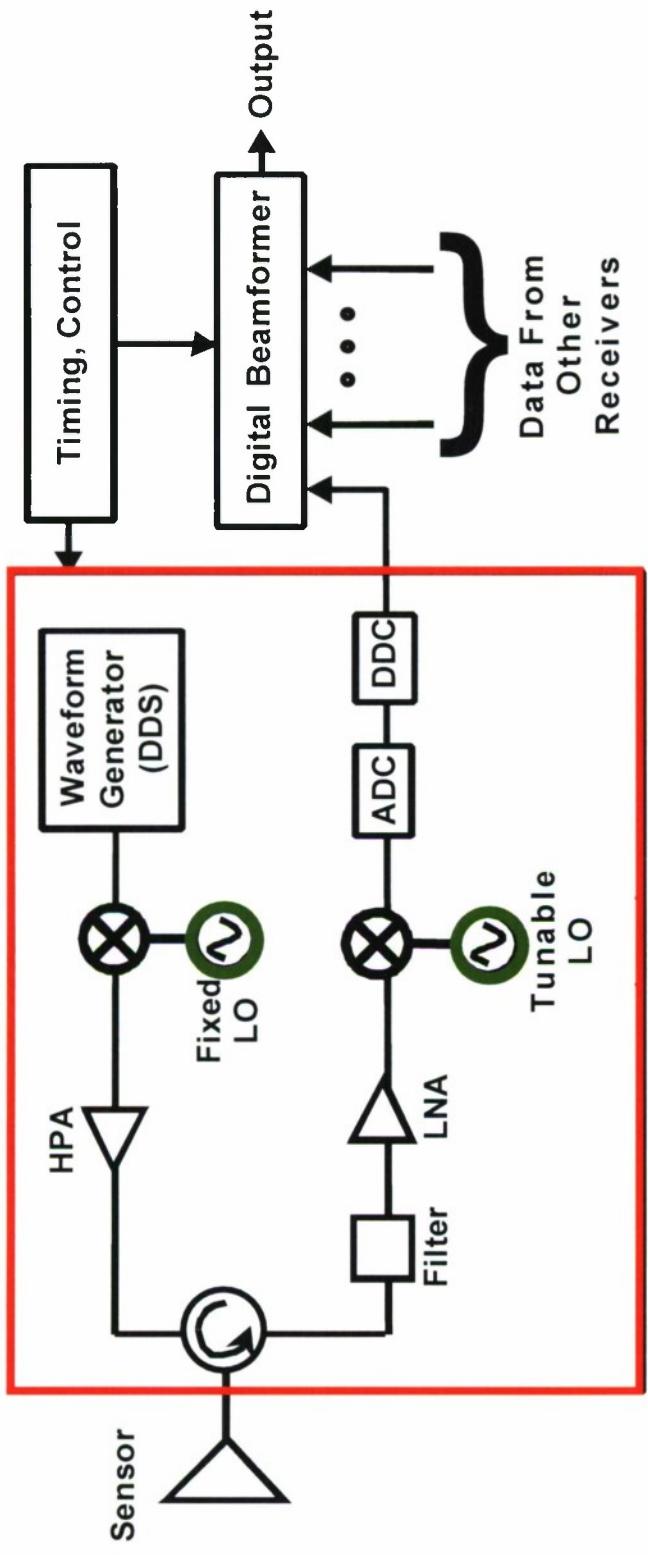
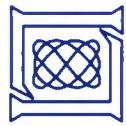
Some Types of Nonlinearity in a Typical Digital T/R Module



- Phase Truncation (DDSSs)

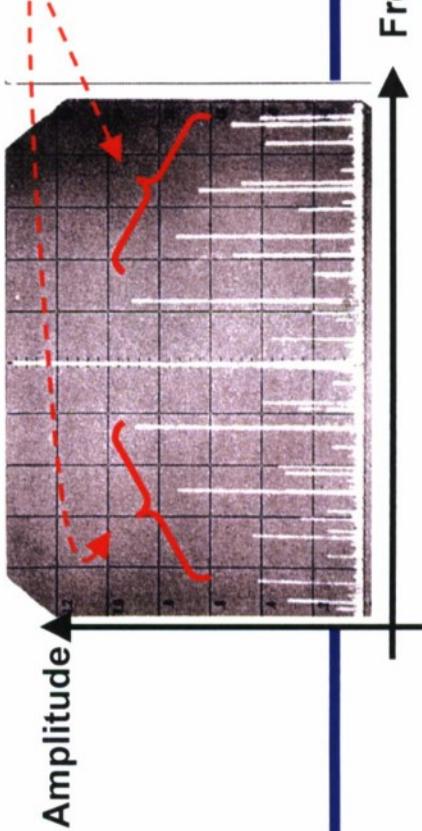


Some Types of Nonlinearity in a Typical Digital T/R Module

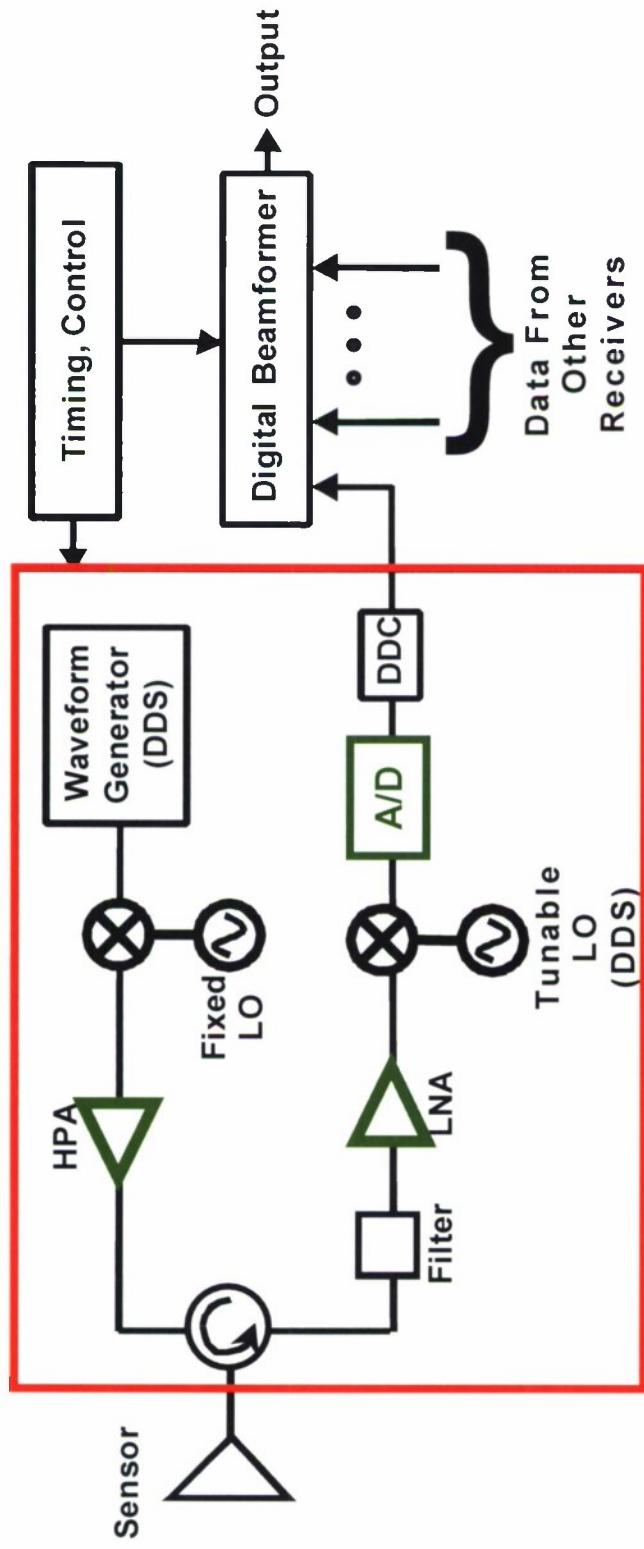
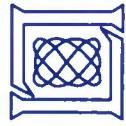


- **M x N Products (Mixers)**

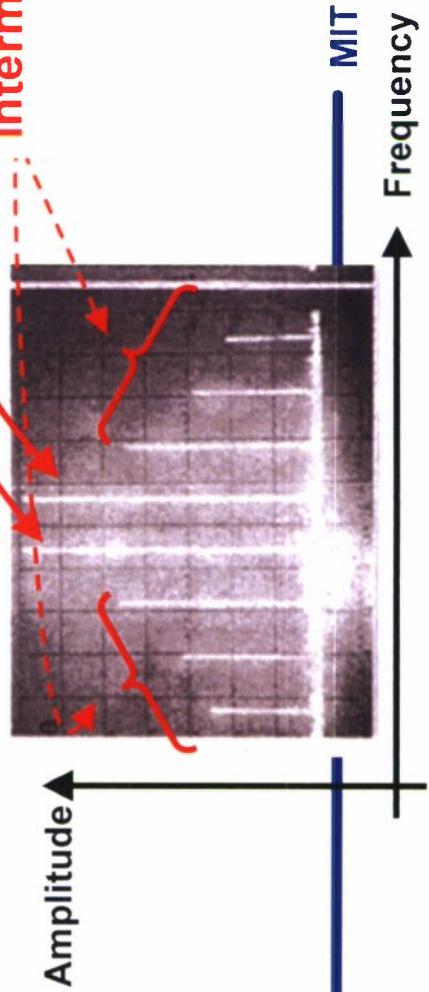
intermodulation of tone
With LO



Some Types of Nonlinearity in a Typical Digital T/R Module

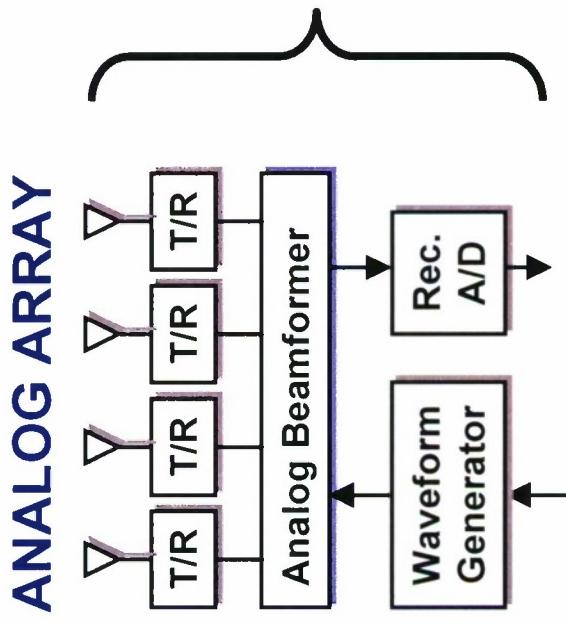


- **Intermodulation (LNA, HPA)**
- **two input tones**
- **intermodulation of tones**



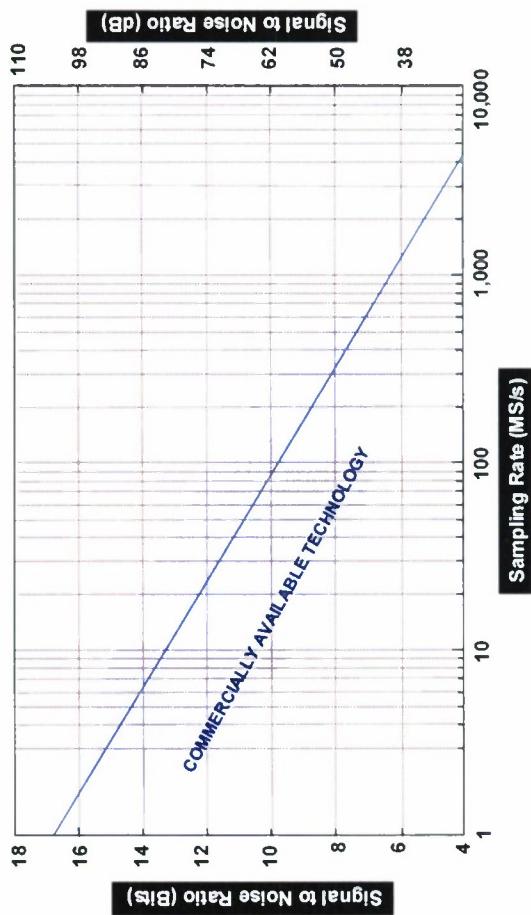


Linearity Requirements in Analog Arrays



$$\text{Dynamic Range} = \frac{P_{\max}}{P_{\min}}$$

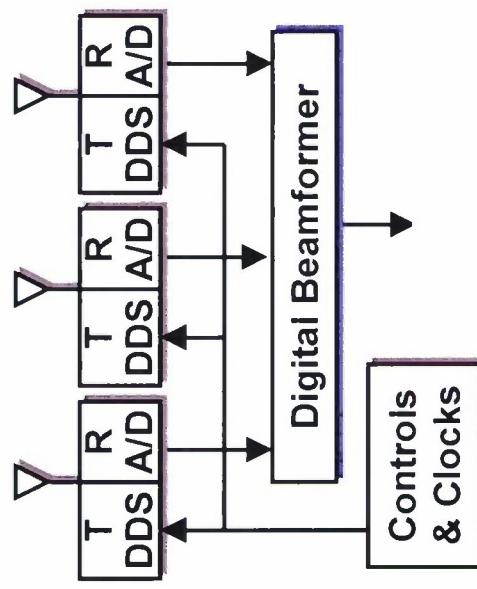
Due to array gain, the dynamic range requirement is usually greatest at the output of beamformer (i.e. the A/D requires the greatest dyn. range)



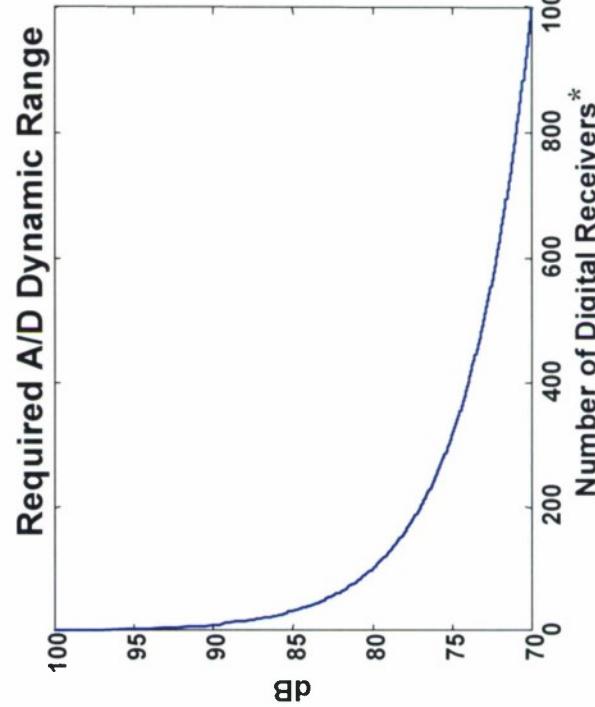


How Digital Arrays “Theoretically” Improve Dynamic Range

DIGITAL ARRAY



- Digital arrays move the waveform generator and A/D closer to the elements
 - Sampled signals are smaller
 - Requirement on A/D, DDS dynamic range is lessened





... but there's one BIG problem

- Even if each receiver's nonlinearities lie below the noise level, the nonlinearities may be correlated spatially (or even coherent).*

If so, the digital beamformer will integrate them (spatially) above the noise, limiting the system's overall dynamic range

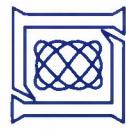
** [We had originally conjected (and proven via mathematic models) that many of these nonlinearities would be correlated. Thanks to the ADC, we've also collected data to show this experimentally ... as you shall see.]*



Outline

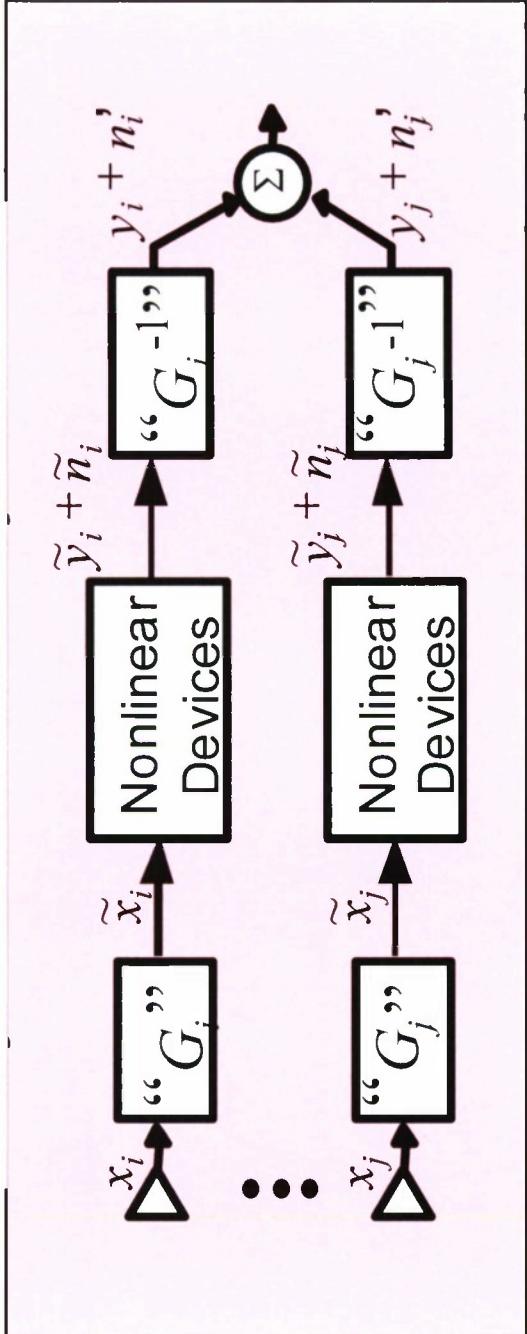
- **Introduction**
- **Importance of linearity and dynamic range**
- **Method for improving dynamic range**
- **Experimental data collection**
- **Summary and future work**





Novel Approach:

-- Decorrelation of Array Nonlinearities --



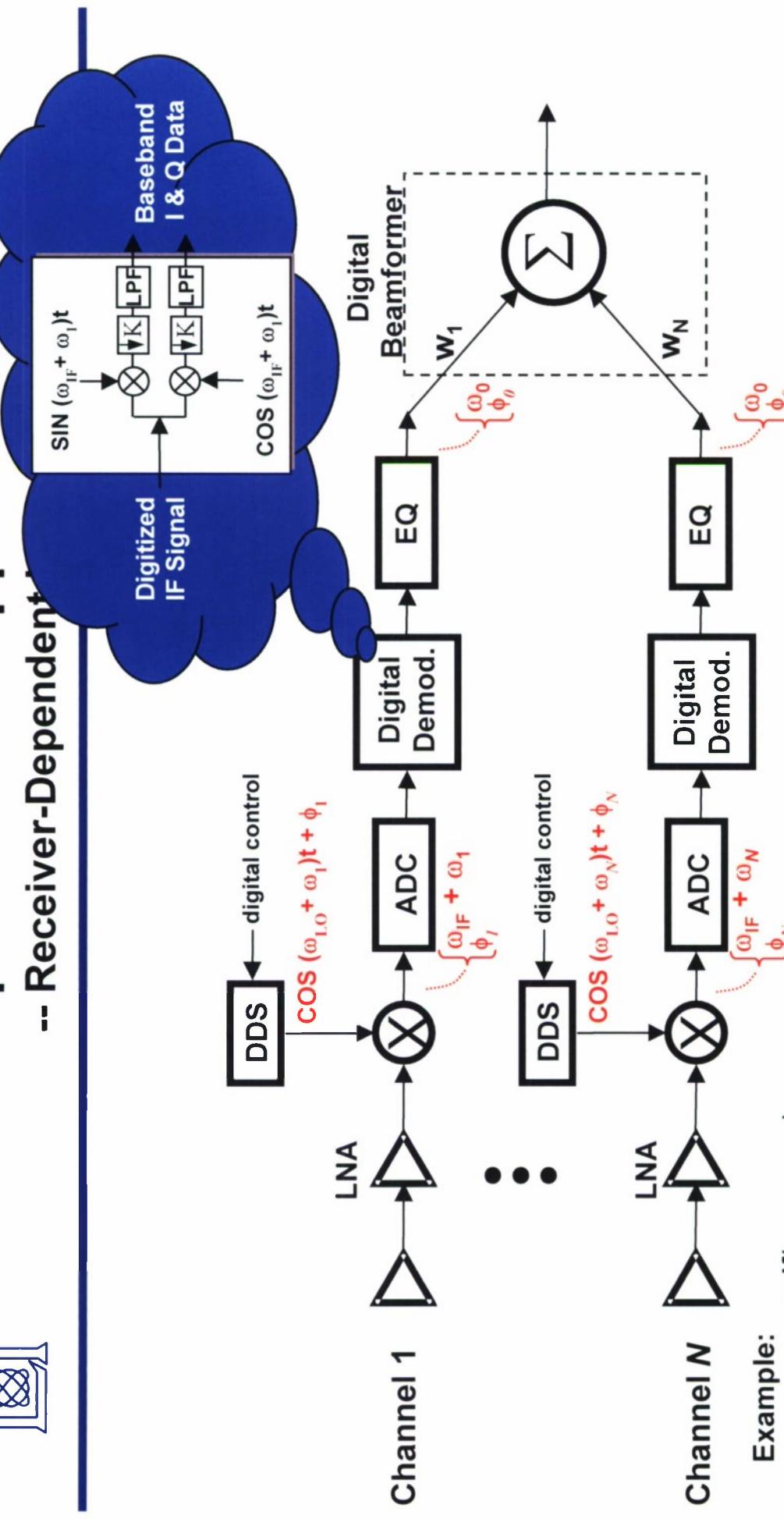
- Input signals are transformed differently at each channel
- “Inverse” transformation later restores the linear component
 - But nonlinear terms still vary from channel to channel, decorrelating them and/or making them sum incoherently.

**NOTE: We do not remove nonlinearities ...
but we do prevent them from being correlated (i.e., no array gain)**



Example of “New Approach”

-- Receiver-Dependent



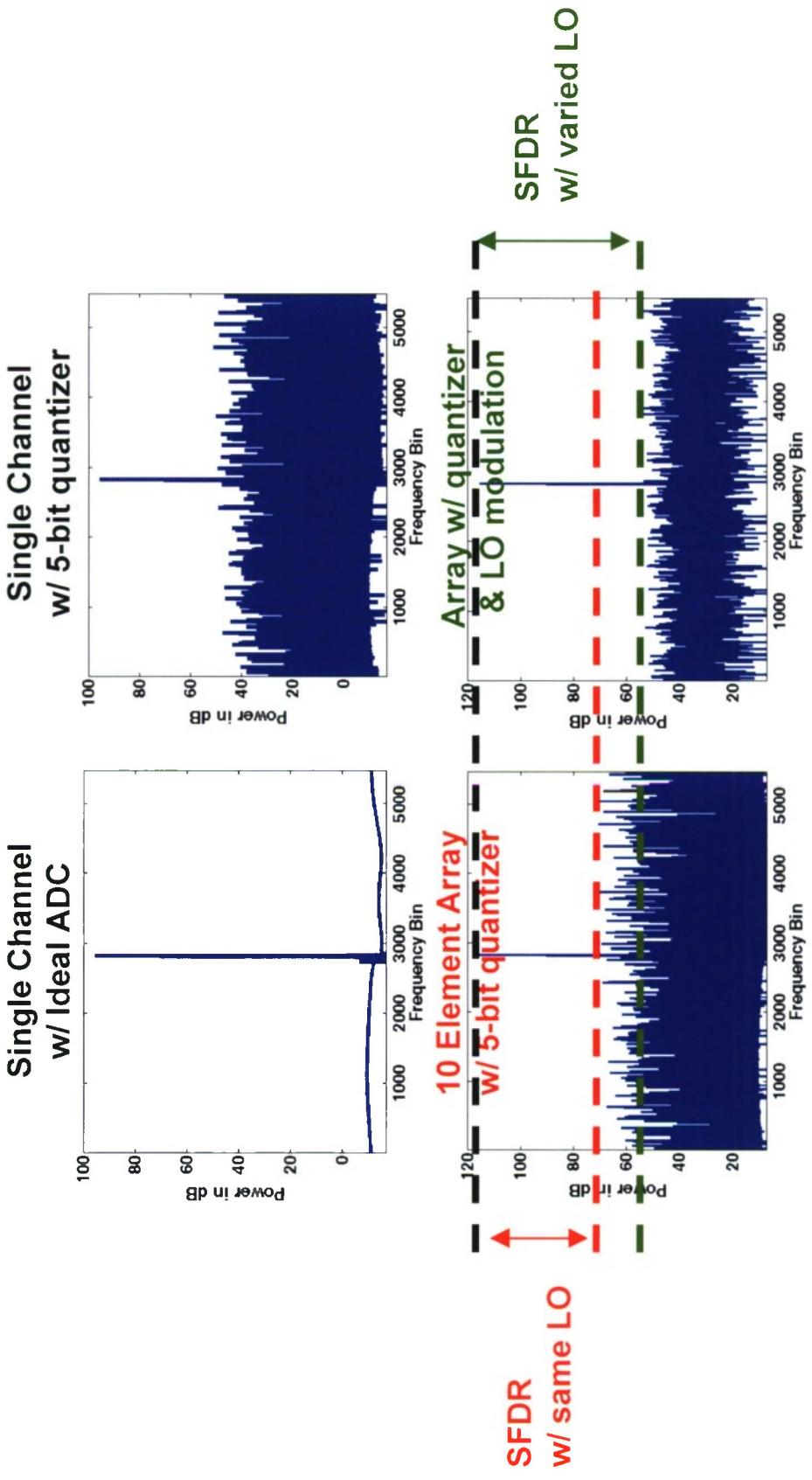
Example:

$$\omega_i \sim \text{unif}(-\omega_{\min}, \omega_{\min})$$
$$\phi_i = 0$$

- **Addresses:** Harmonically related spurs generated by ADC, DDS and DAC, **MxN** products in mixer, and LO and ADC phase noise
 - Varied LO causes different sets of frequencies to be excited

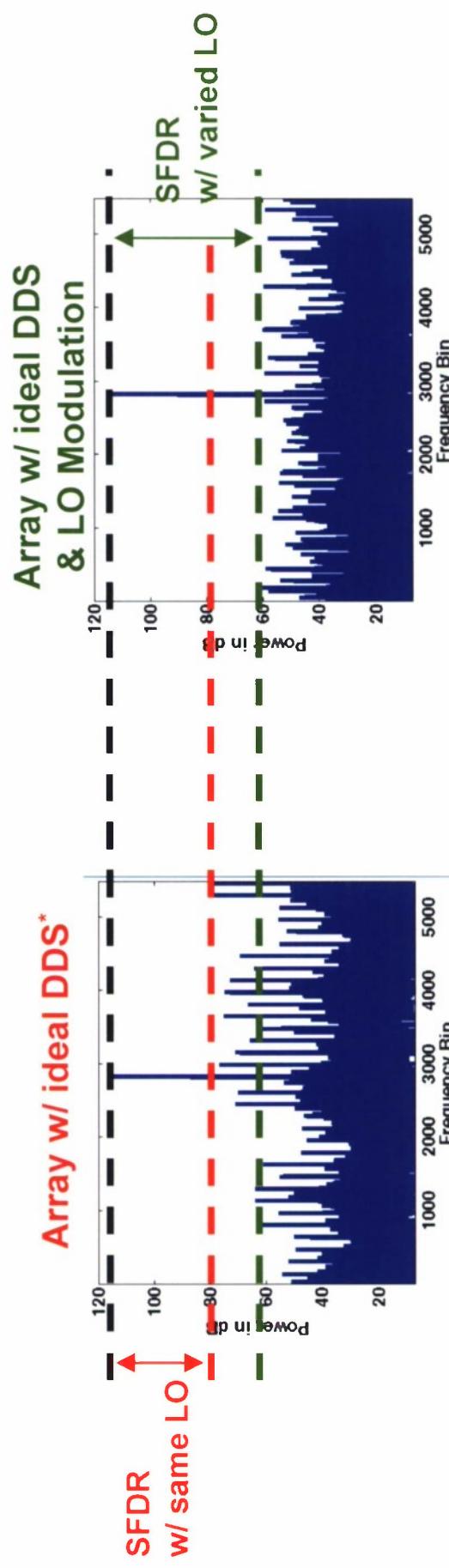
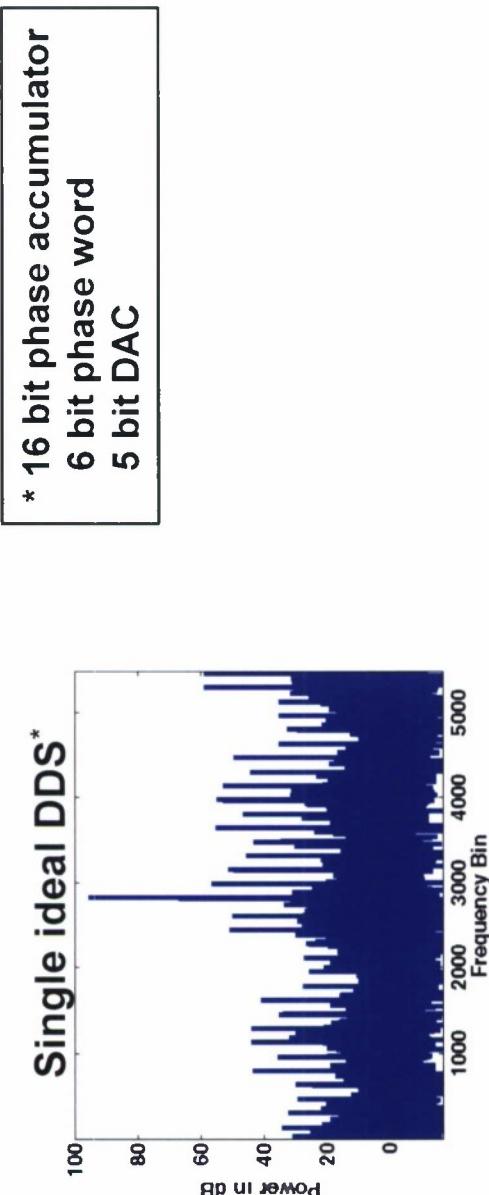


Predicted A/D Nonlinearities with Receiver-Dependent LO



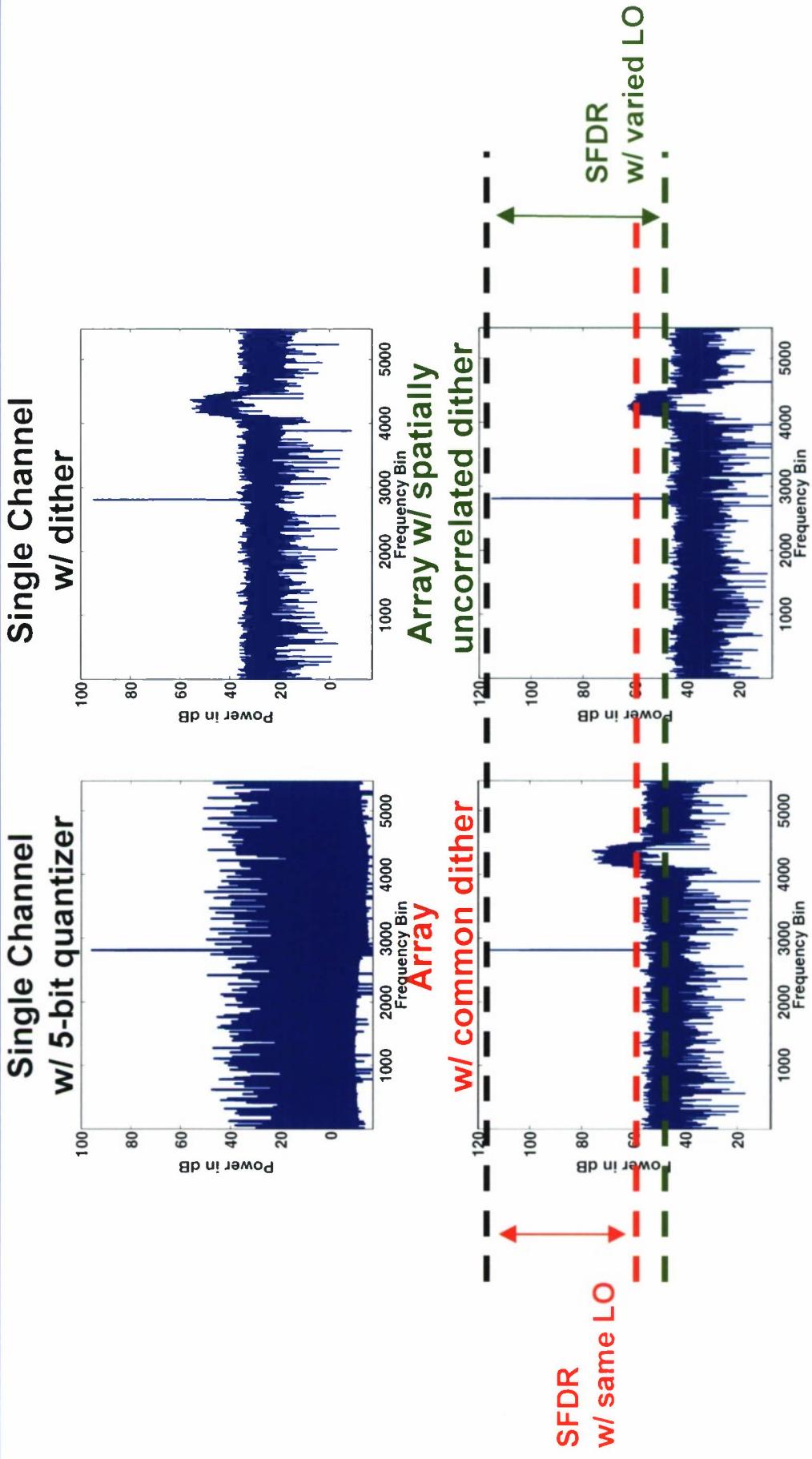
- Randomized LO modulation makes spurs appear uncorrelated, thus avoiding array gain

Predicted DDS Nonlinearities with Receiver-Dependent LO





Example 2: Receiver-Dependent Dither

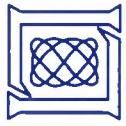


- Randomized dither makes spurs appear uncorrelated, thus avoiding array gain (ADCs and DACs only)



Other Types of Distortion

- $M \times N$ products and harmonics can often be addressed by phase transformations
- Intermods from well-separated sources can be addressed by nonlinear-phase all-pass filtering and subsequent equalization
- Closer-in intermods are not addressed directly
 - However, by first mitigating other complex nonlinearities as described (so that they're below the noise), a relatively simple equalizer can be used to remove the remaining close-in intermods (example given later).
- Method above can be applied on transmit in analogous fashion



Outline

- **Introduction**
- **Importance of linearity and dynamic range**
- **Method for improving dynamic range**
- **Experimental data collection**
- **Future work and summary**





Outline

- **Introduction**
- **Importance of linearity and dynamic range**
- **Method for improving dynamic range**
- **Experimental data collection**
 - Objectives
 - Mitigation of correlated nonlinearities
 - Uncorrelated nonlinearity
- **Future work and summary**





ADC Testbed Objectives

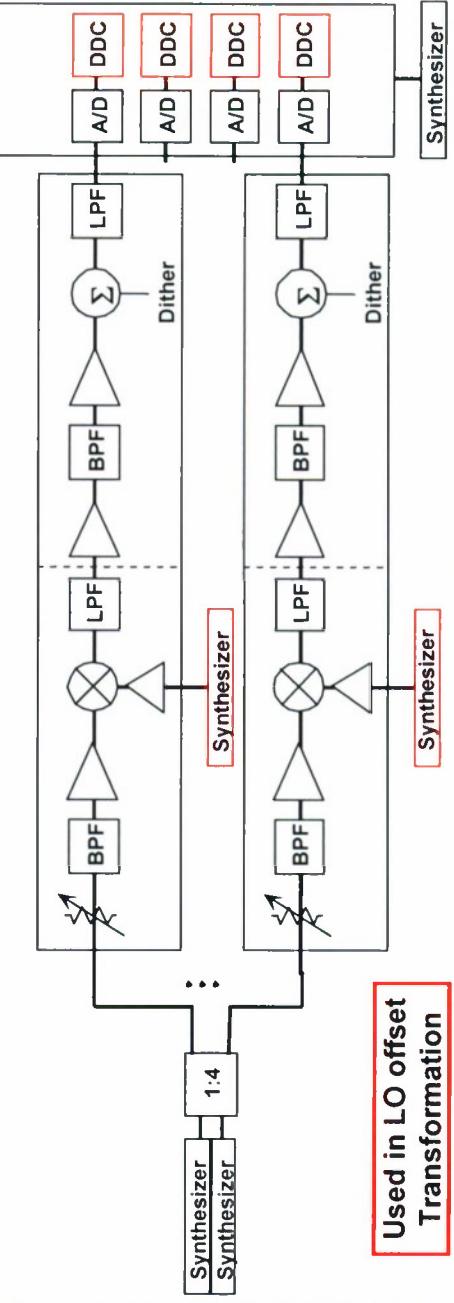
Goals

- Experiment fit within time and hardware constraints
- Only four channels available
- Linear Array Upgrade program had 2-week window of availability for hardware
- Test experimentally whether or not distortions add in beamformer
- Apply mitigation technique to these cases

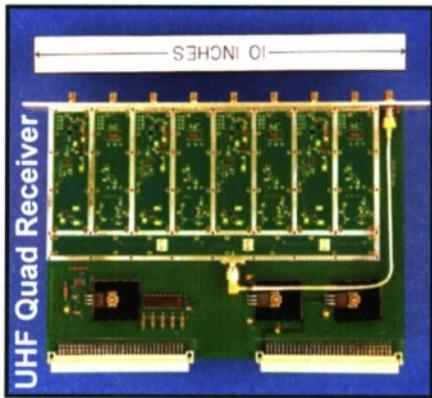
Hardware

- Group 101 Quad UHF receiver
- Pentek 4290 DSP with pair of 6216 dual digital receiver
 - NCO in DDC has < 1 Hz resolution
 - 12 Bit 65 MSPS ADC
- Graychip Digital Down-Converter (DDC) GC1016
- VME Downconverter

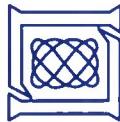
Pentek 4-Channel Development System



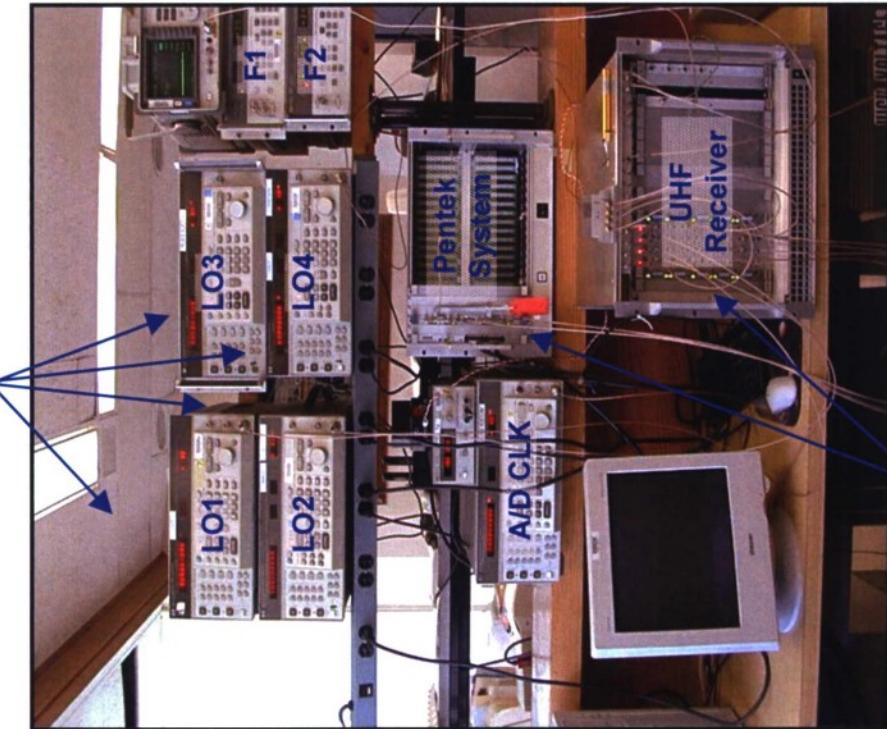
Used in LO offset Transformation



Experiments



Distributed LO System



Case #	Spur investigated	F1 (MHz)	F2 (MHz)	LO1 (MHz)	LO2 (MHz)	LO3 (MHz)	LO4 (MHz)
1	a -2 x 2 product	422	458	493	—	—	—
	b -2 x 2 product	422	458	493	493+ δ 1	493+2* δ 1	493+3* δ 1
	c -2 x 2 product	422	458	493	493+ δ 2	493+2* δ 2	493+3* δ 2
2	a -3 x 3 product	420	468	491	—	—	—
	b -3 x 3 product	420	468	491	491+ δ 1	491+2* δ 1	491+3* δ 1
	c -3 x 3 product	420	468	491	491+ δ 2	491+2* δ 2	491+3* δ 2
3	A/D Harmonic without Dither	75					
	a Receiver and A/D Harmonics	435		505	—	—	—
4	b Receiver and A/D Harmonics	435		505	505+ δ 1	505+2* δ 1	505+3* δ 1
	c Receiver and A/D Harmonics	435		505	505+ δ 2	505+2* δ 2	505+3* δ 2
	a Intermodulation	435	435	505	—	—	—
5	b Intermodulation	435	435	505	505+ δ 1	505+2* δ 1	505+3* δ 1
	c Intermodulation	435	435	505	505+ δ 2	505+2* δ 2	505+3* δ 2
	d Intermodulation	435	435	505	—	—	—
6	e Intermodulation	435	435	505	—	—	—
	a Receiver noise			505	—	—	—
	b Receiver noise			505	505+ δ 1	505+2* δ 1	505+3* δ 1
7	c Receiver noise			505	505+ δ 2	505+2* δ 2	505+3* δ 2
	a A/D noise						

Digital Receivers



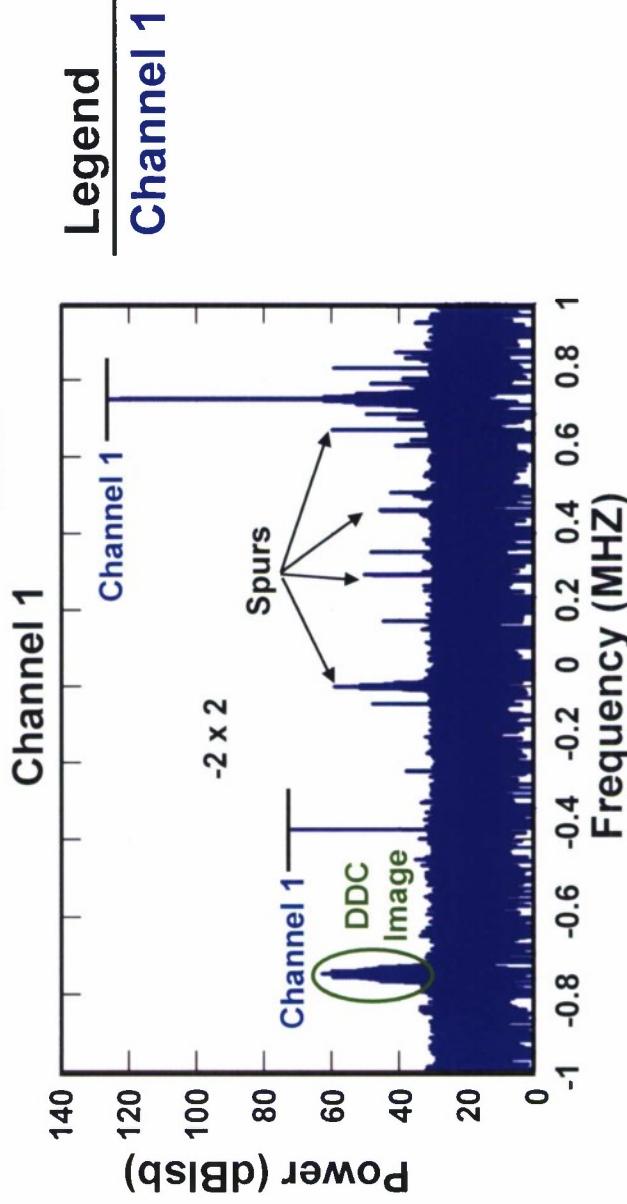
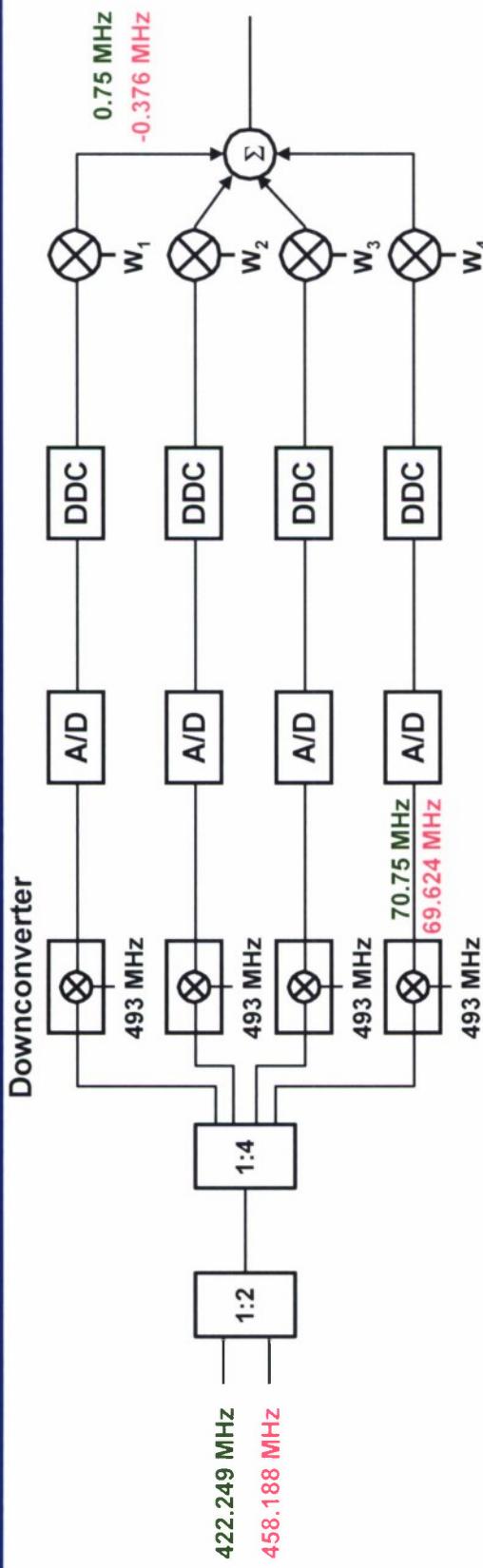
Outline

- **Introduction**
- **Importance of linearity and dynamic range**
- **Method for improving dynamic range**
- **Experimental data collection**
 - Objectives
 - Mitigation of correlated nonlinearities
 - Uncorrelated nonlinearity
- **Future work and summary**



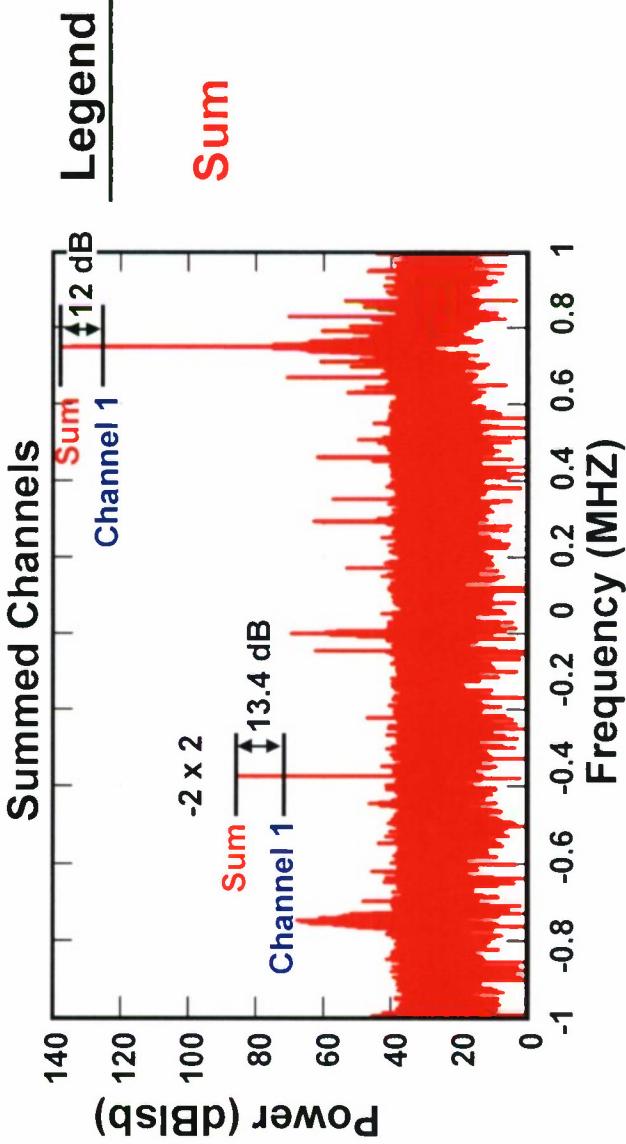
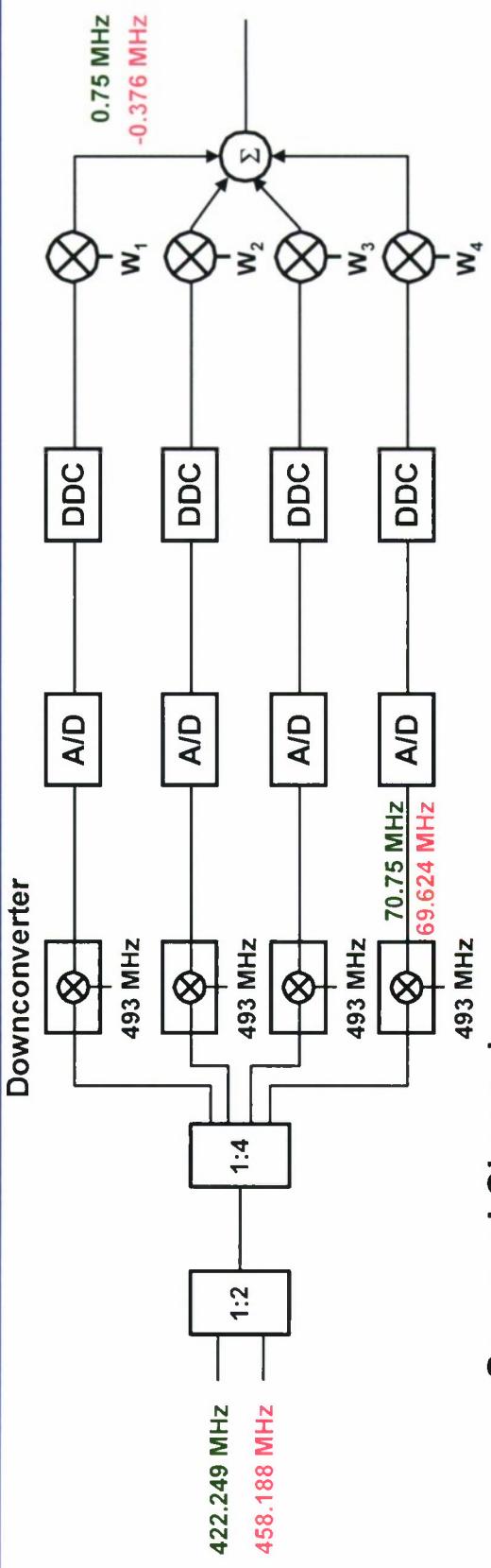


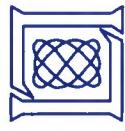
-2 x 2 Mixer Intermodulation Product Correlation



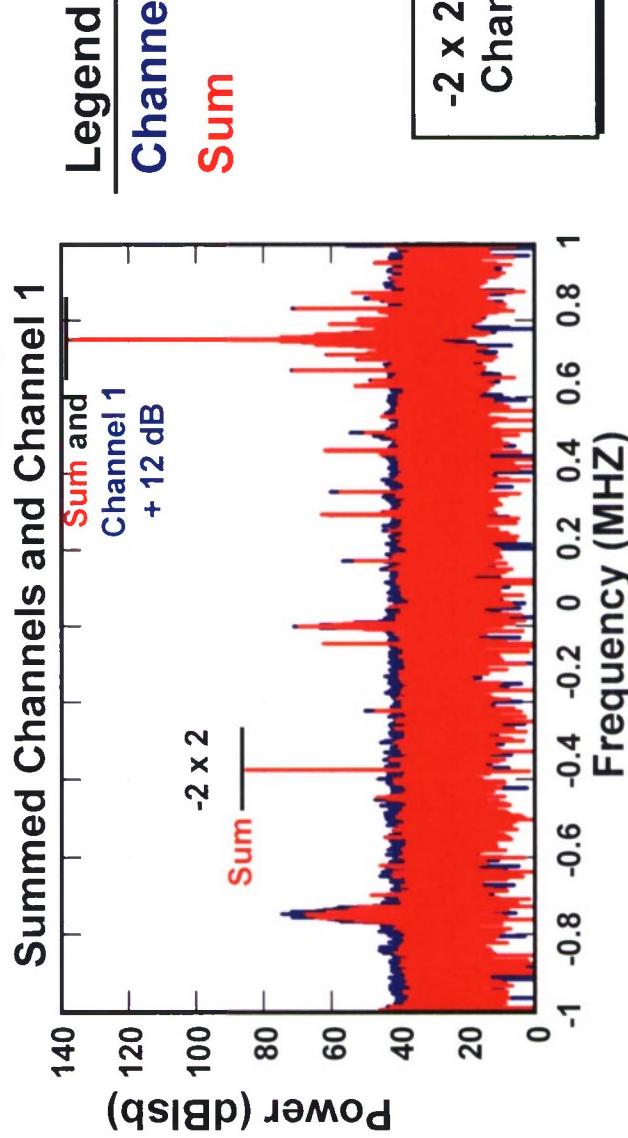
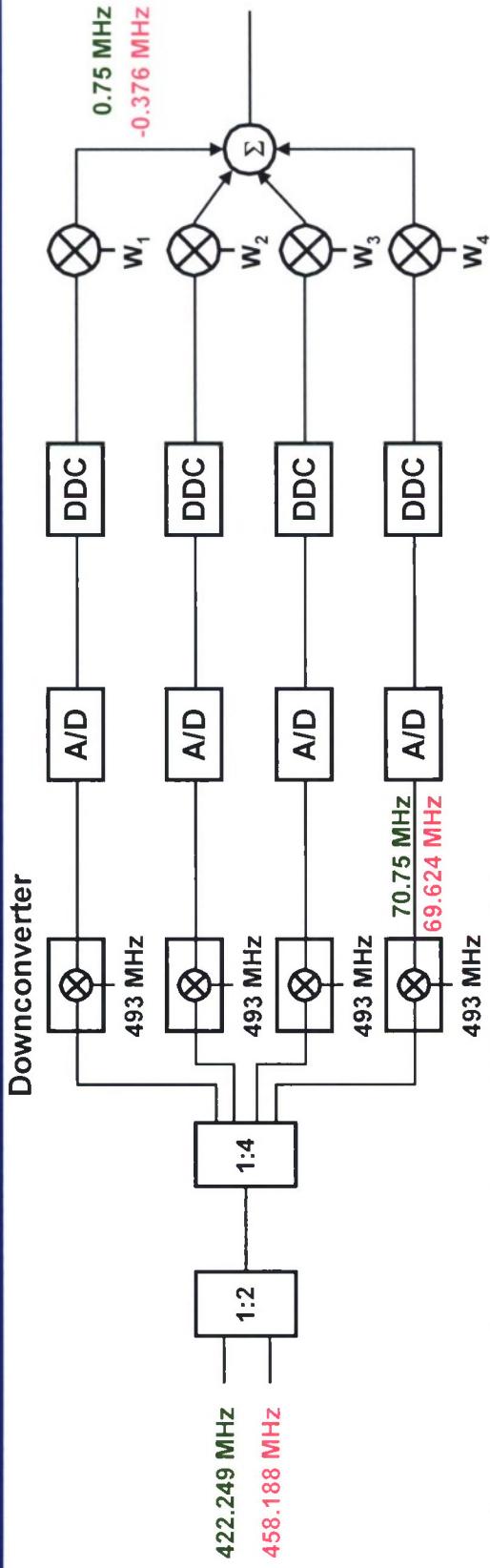


-2 x 2 Mixer Intermodulation Product Correlation



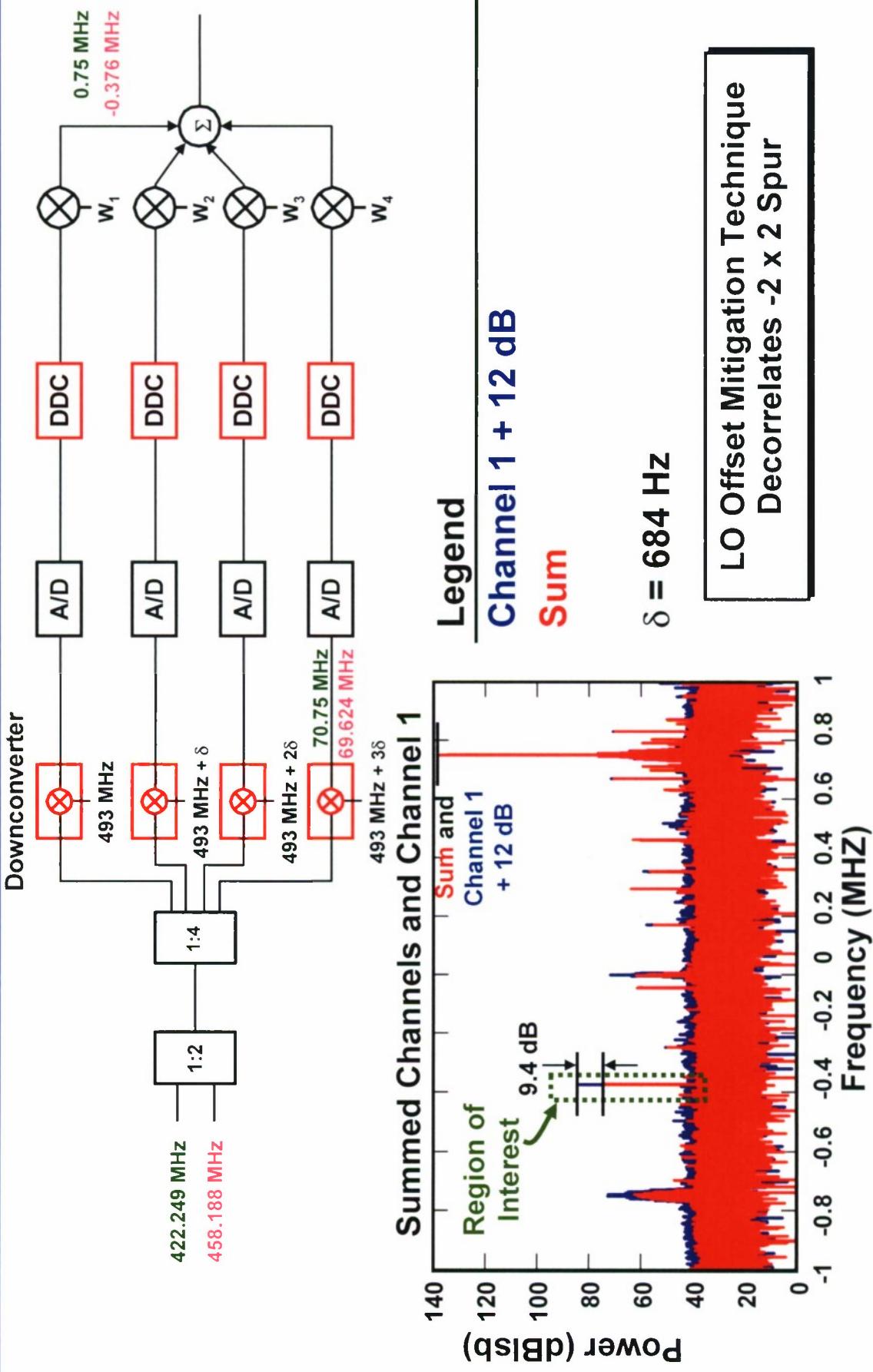


-2 x 2 Mixer Intermodulation Product Correlation



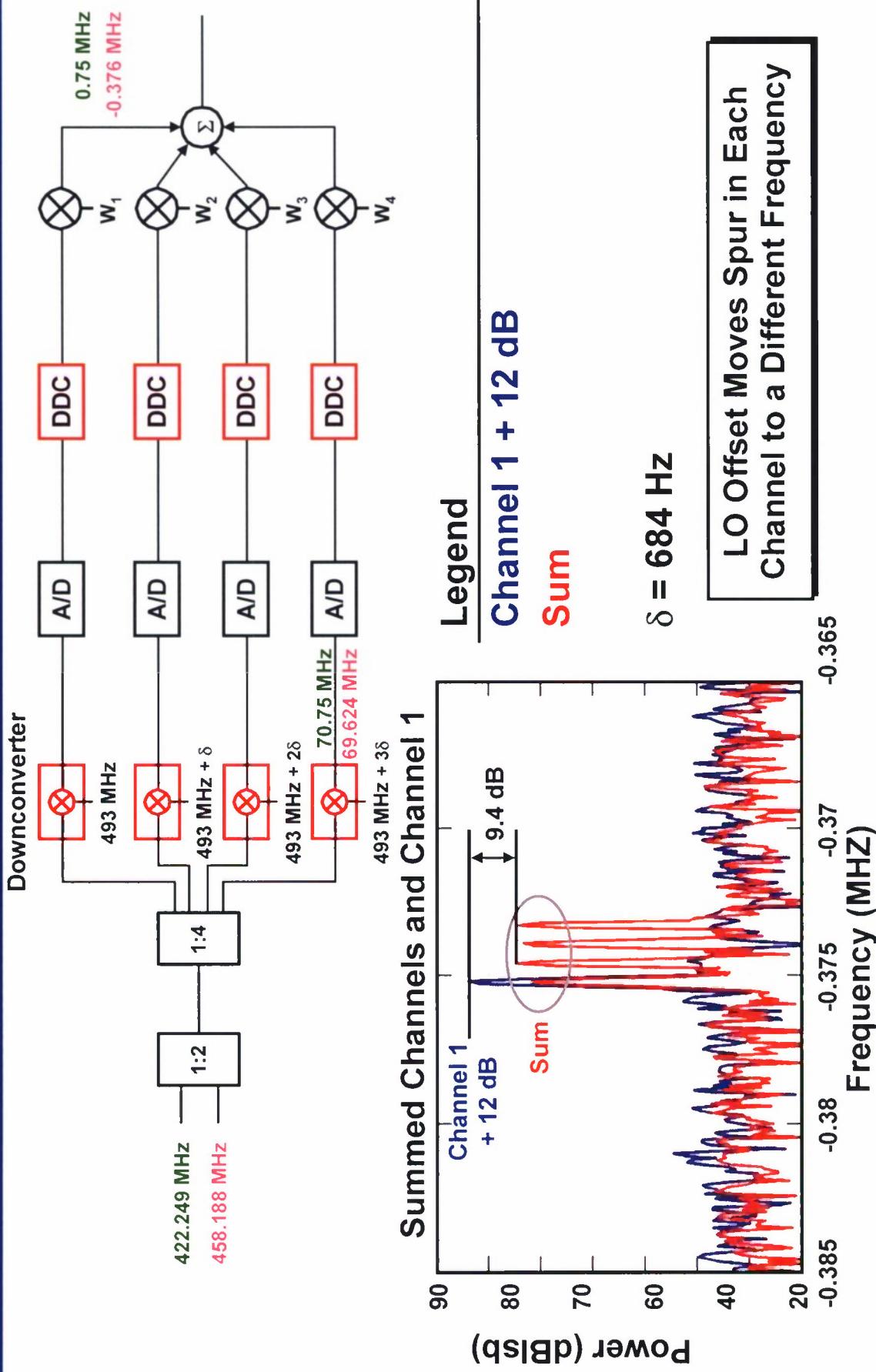


LO Offset Mitigation Applied to -2 x 2 Mixer Intermodulation Product Correlation

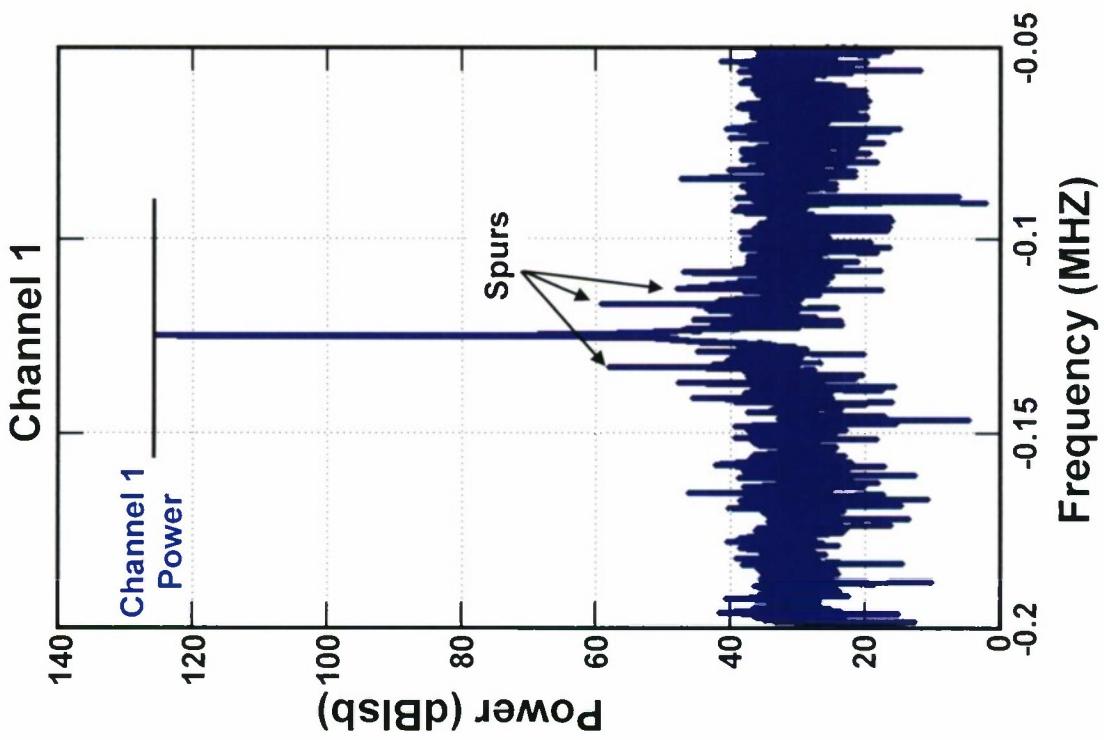
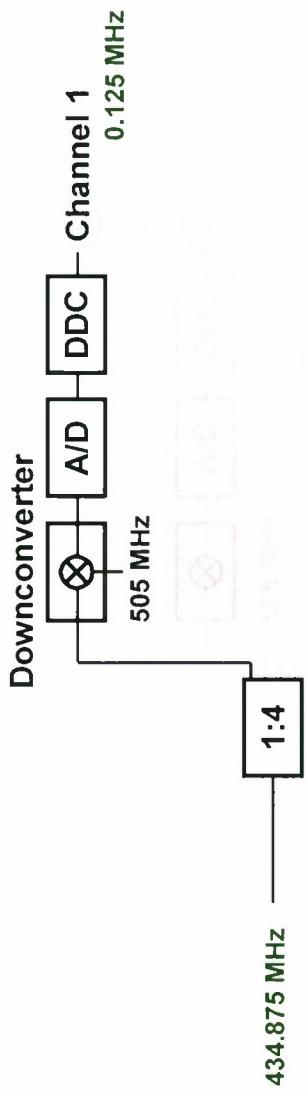




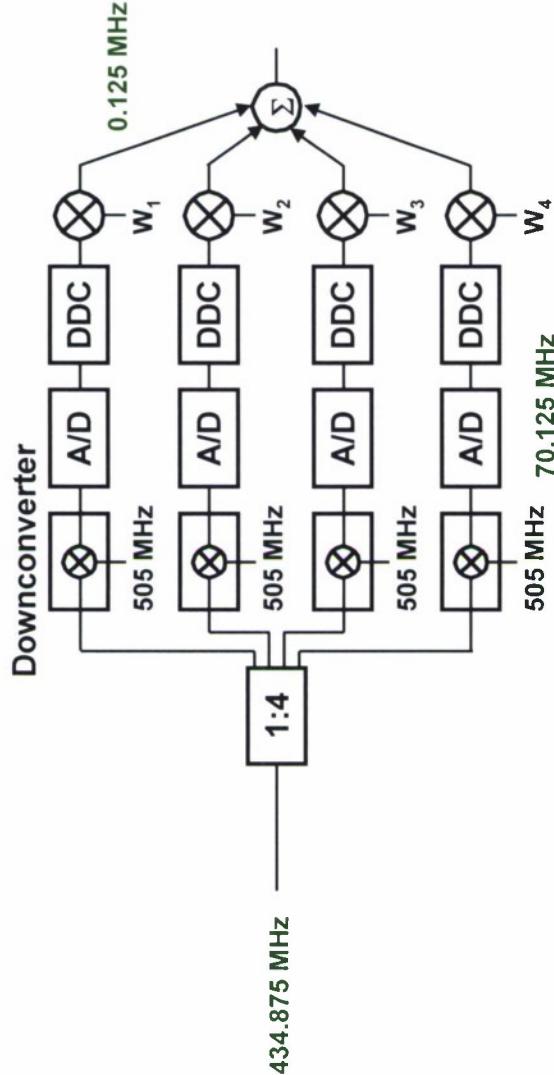
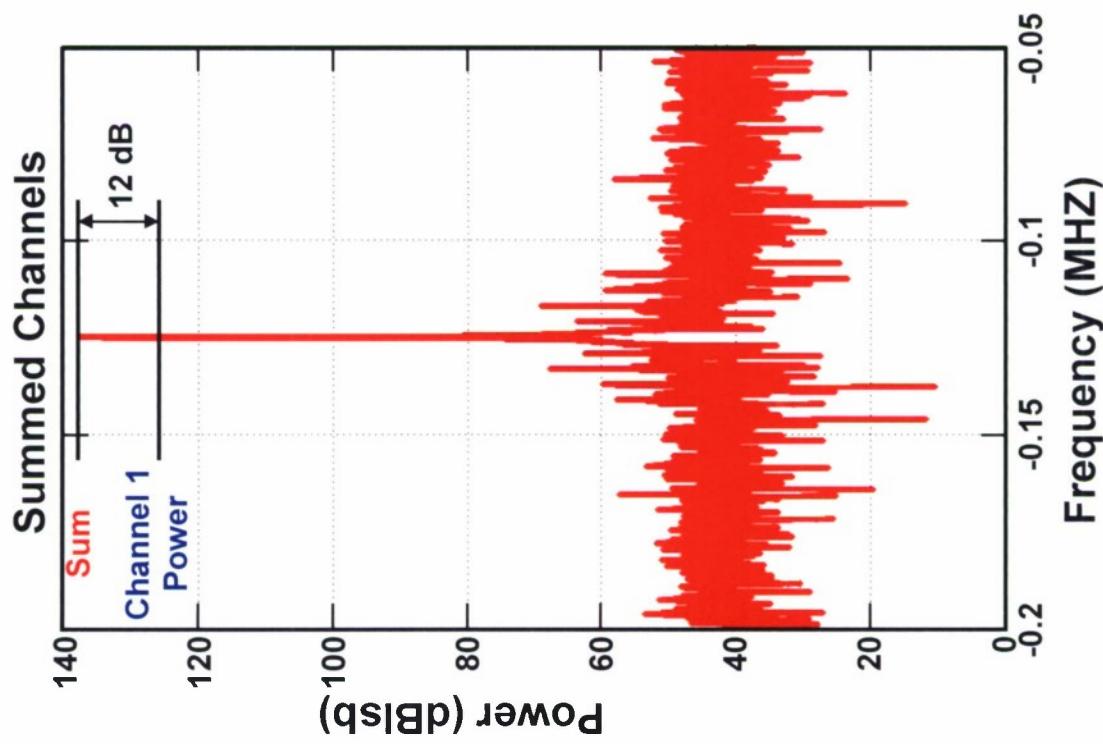
LO Offset Mitigation Applied to -2 x 2 Mixer Intermodulation Product Correlation



Noise Correlation



Noise Correlation

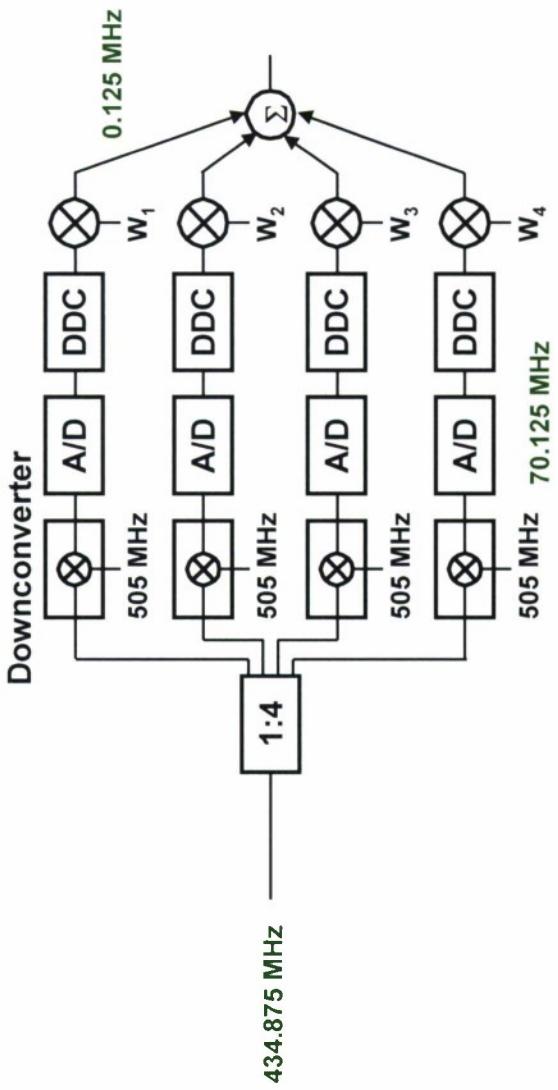
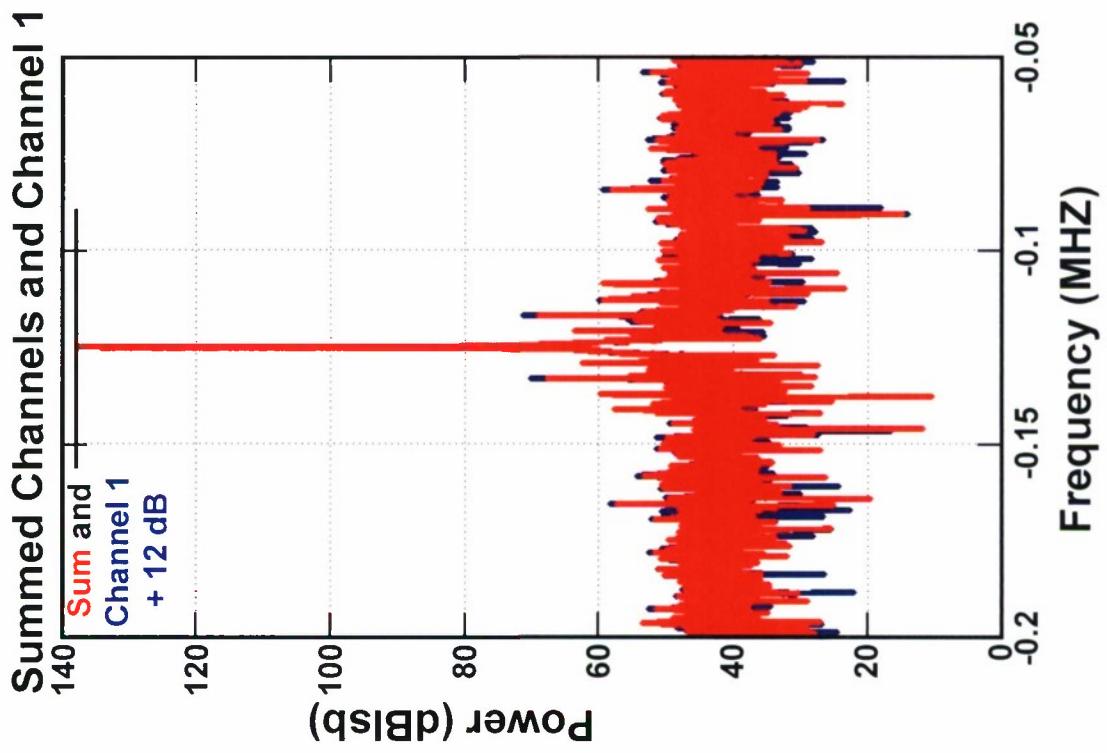


Legend

Sum of 4 Channels



Noise Correlation



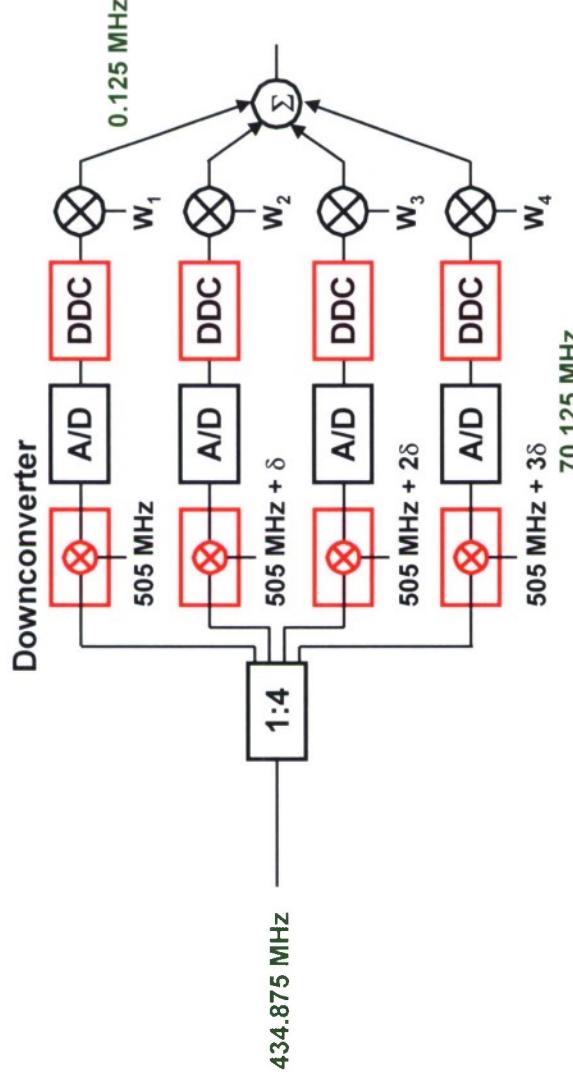
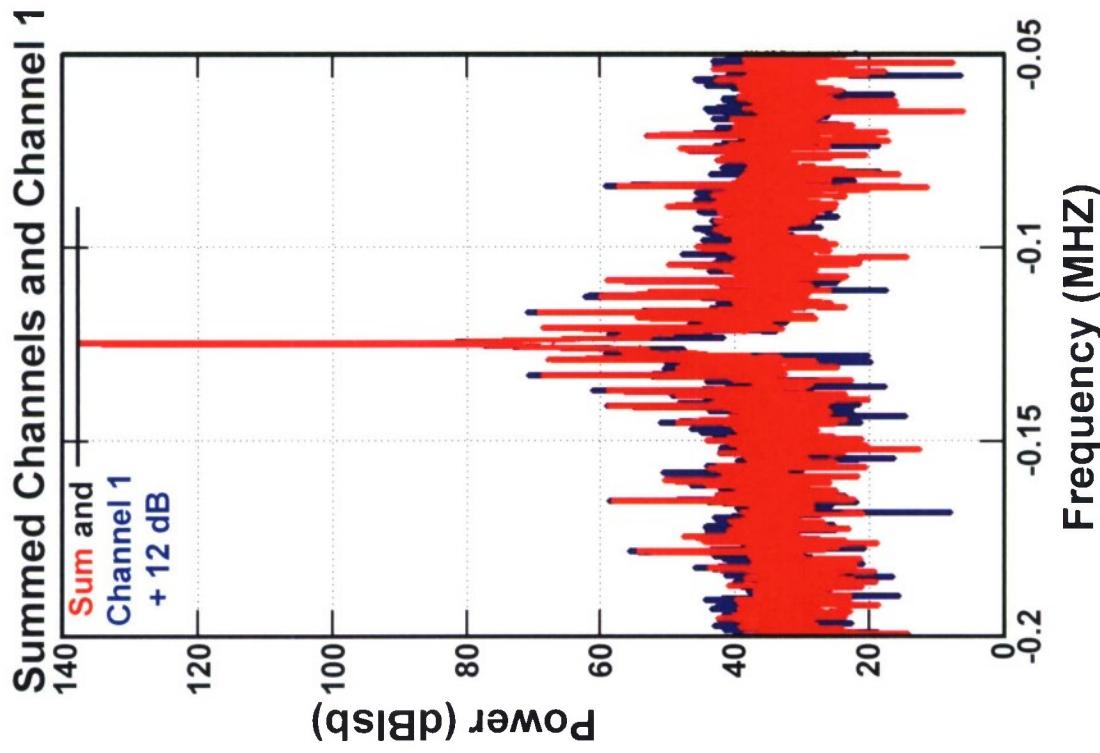
Legend

Channel 1 + 12 dB
Sum of 4 Channels

**SNR of 4 Channels only 0.8 dB higher
(vs 6 dB expected)**
⇒ Phase Noise is correlated!



LO Offset Mitigation Applied to Noise Correlation Data



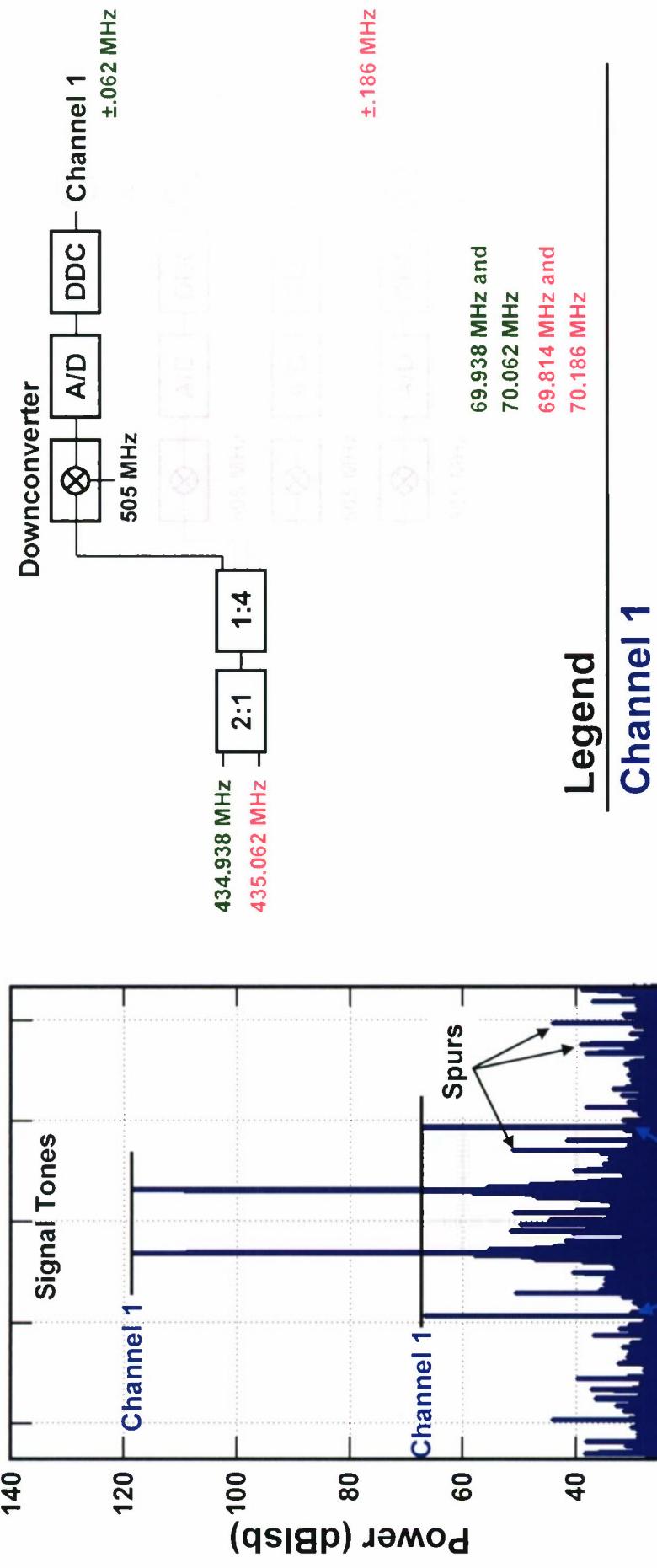
Legend

Channel 1 + 12 dB
Sum of 4 Channels

SNR Gain on Beamforming is Now
4.1 dB (vs 0.8 dB)



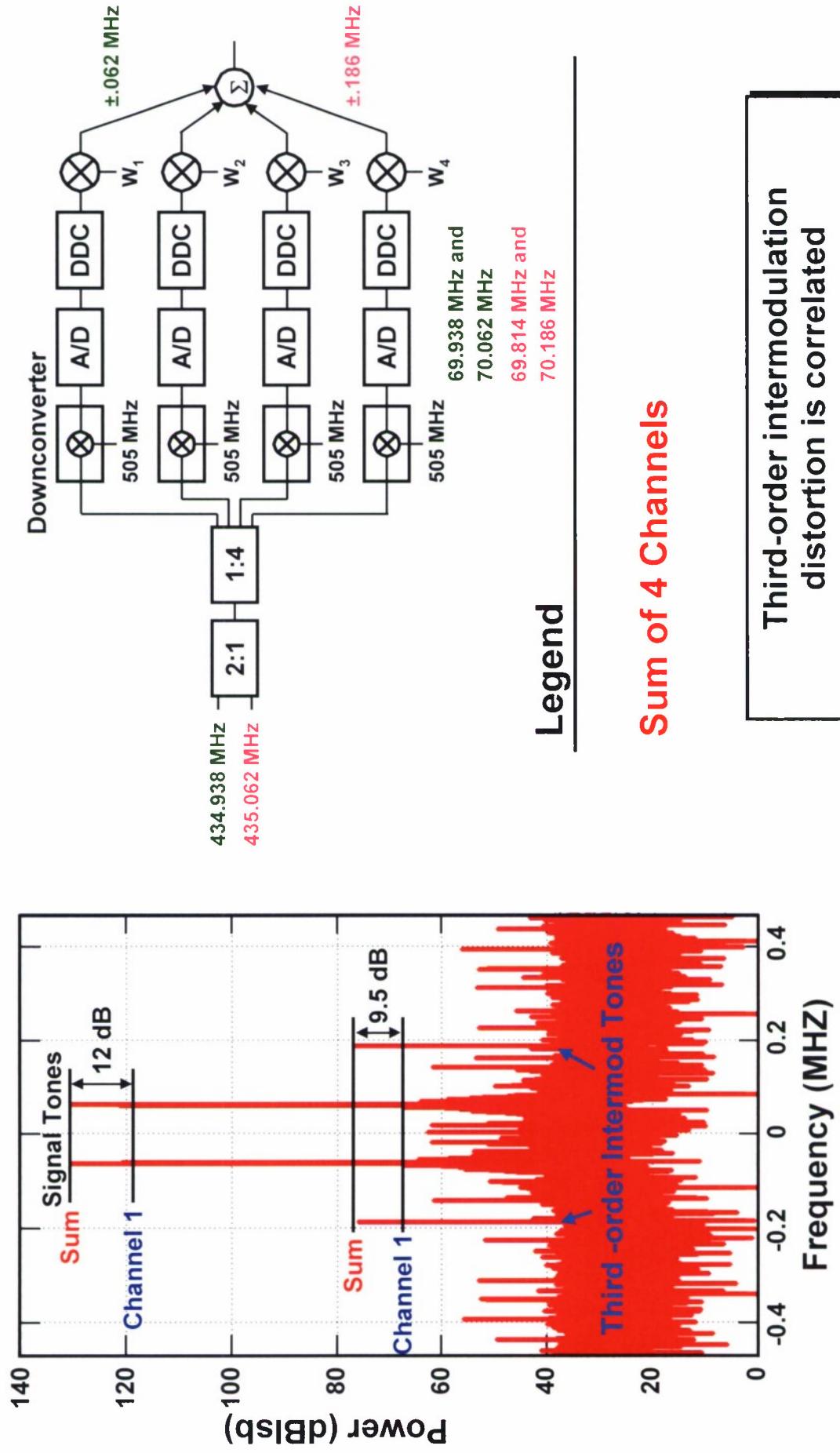
Third-order Intermodulation Correlation

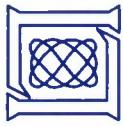


020604 Page # 34
MTT 2002

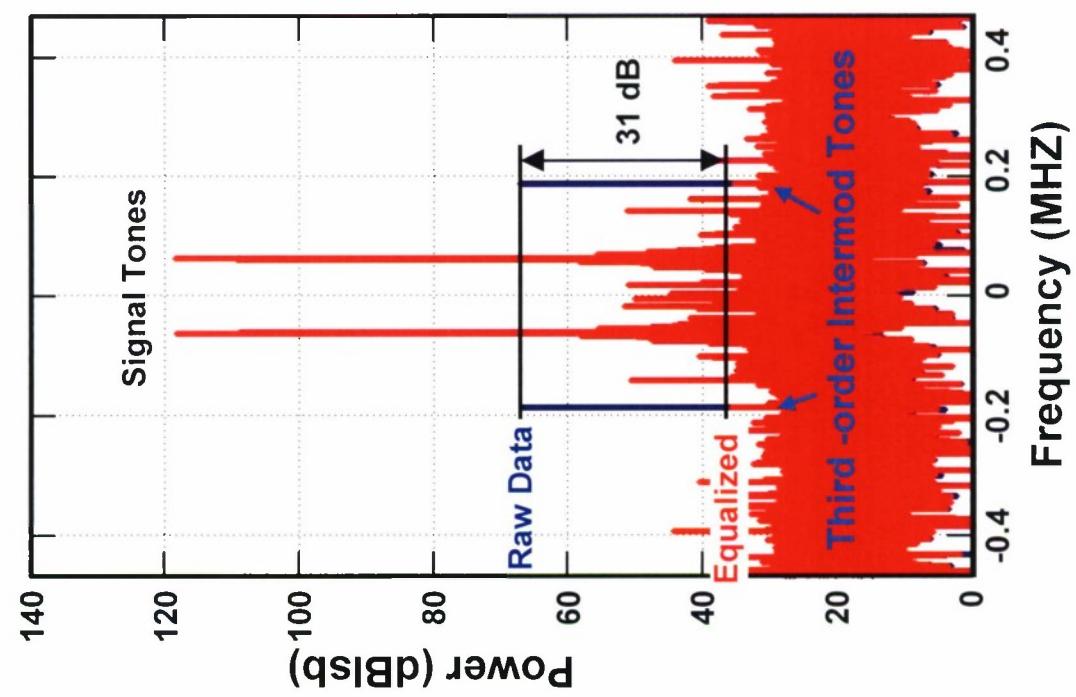


Third-order Intermodulation Correlation

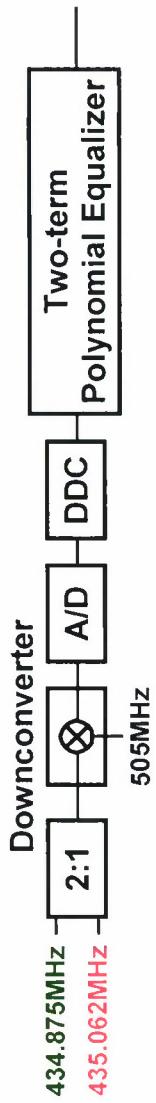




Third-order Intermodulation Mitigation



Approach: Mitigate most undesired terms by decorrelating them, as described above. Then use a simple, low-order nonlinear equalizer to cancel the rest.



Legend

Raw Data

Equalized

Nonlinear Equalization Reduces Third-order Intermodulations by 31 dB

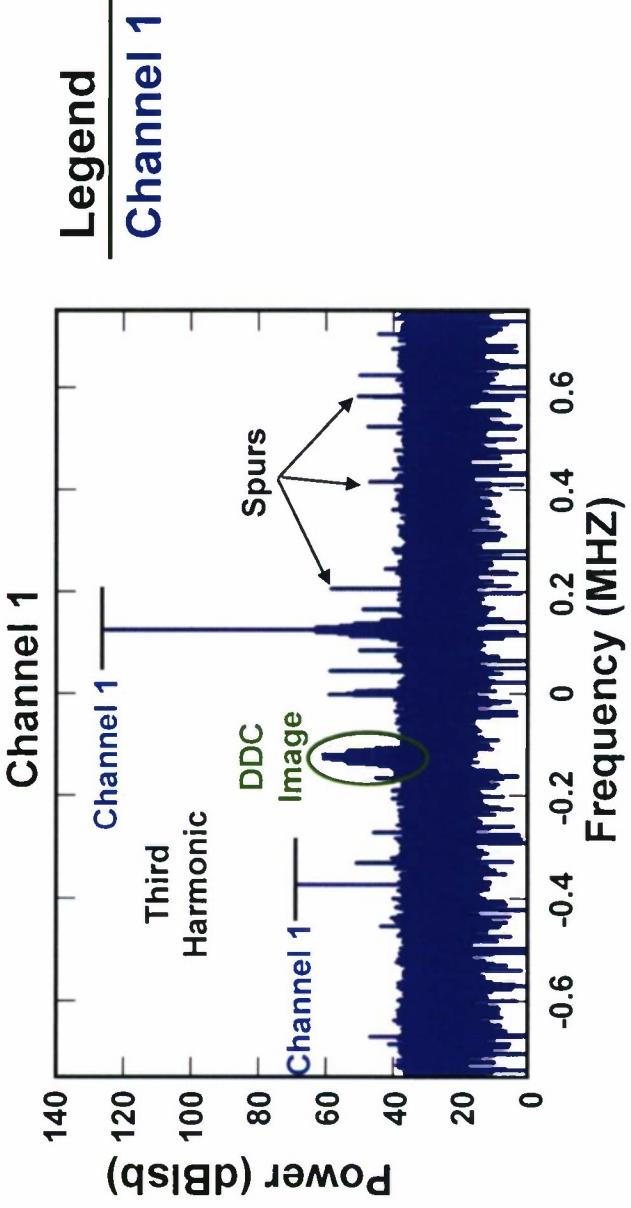
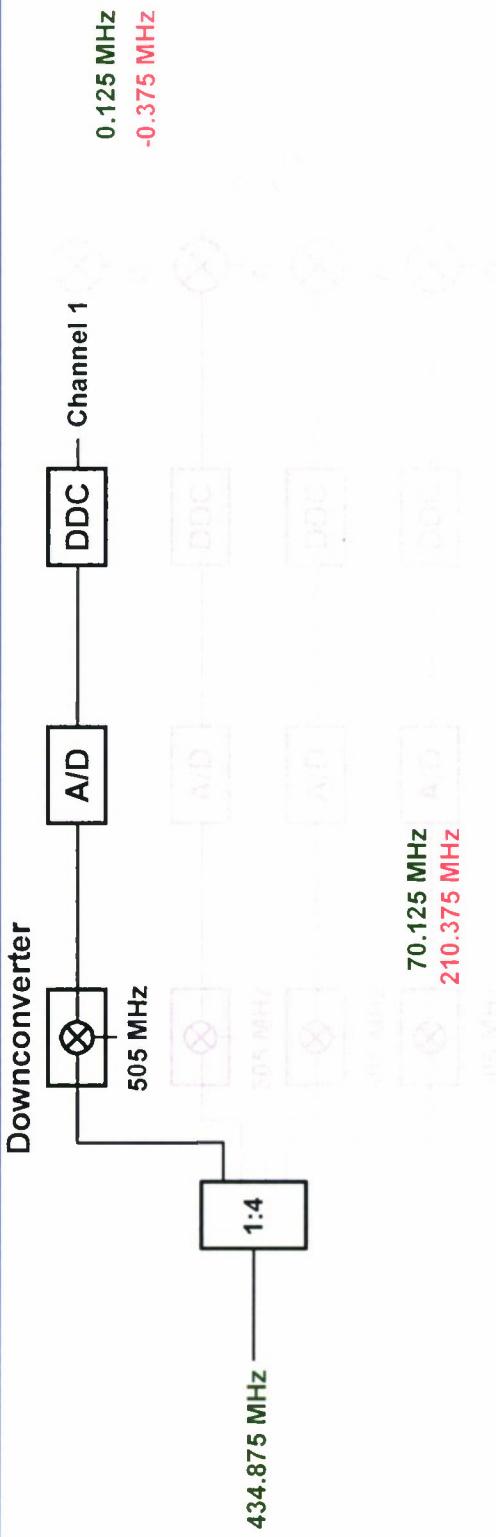


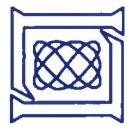
Outline

- **Introduction**
- **Importance of linearity and dynamic range**
- **Method for improving dynamic range**
- **Experimental data collection**
 - Objectives
 - Mitigation of correlated nonlinearities
 - Uncorrelated nonlinearity
- **Future work and summary**

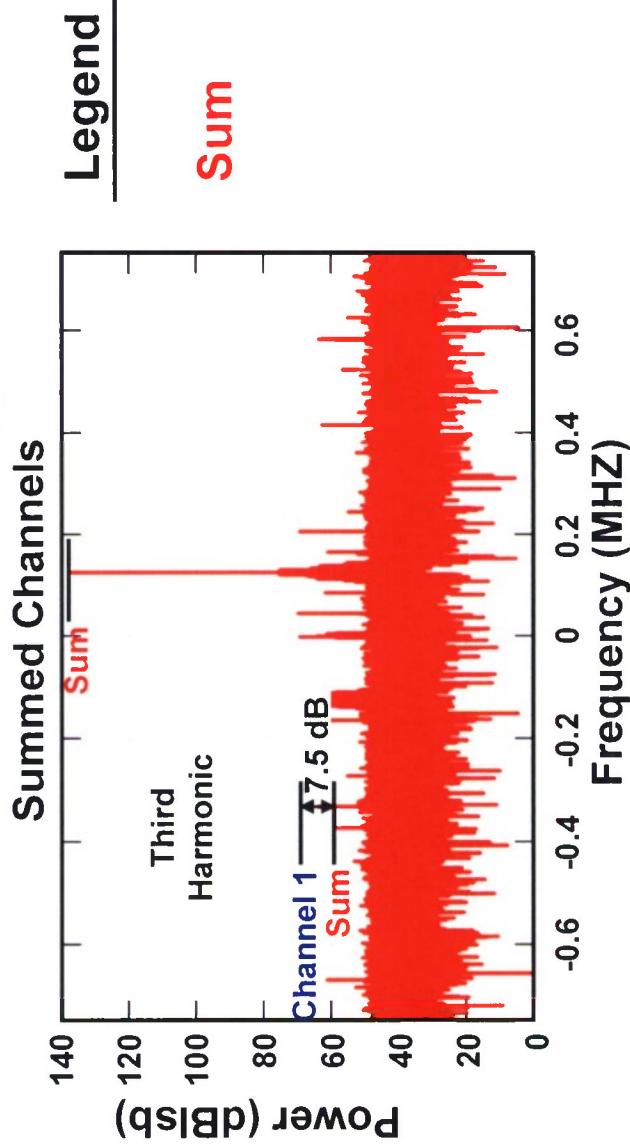
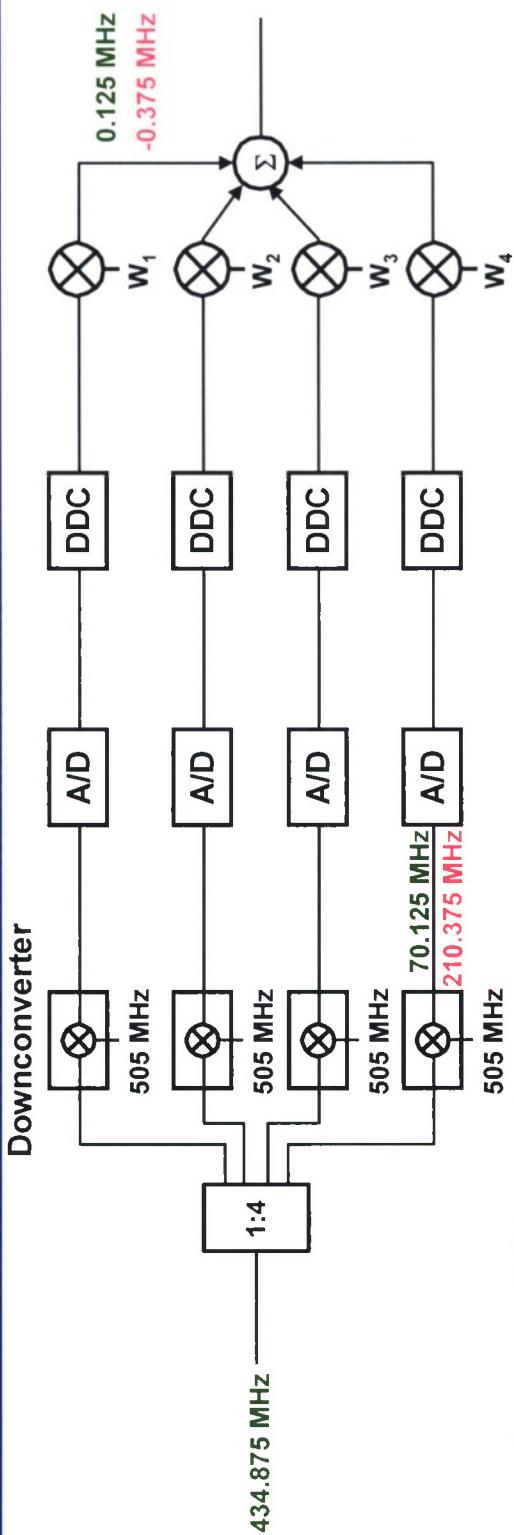


Third Harmonic Correlation



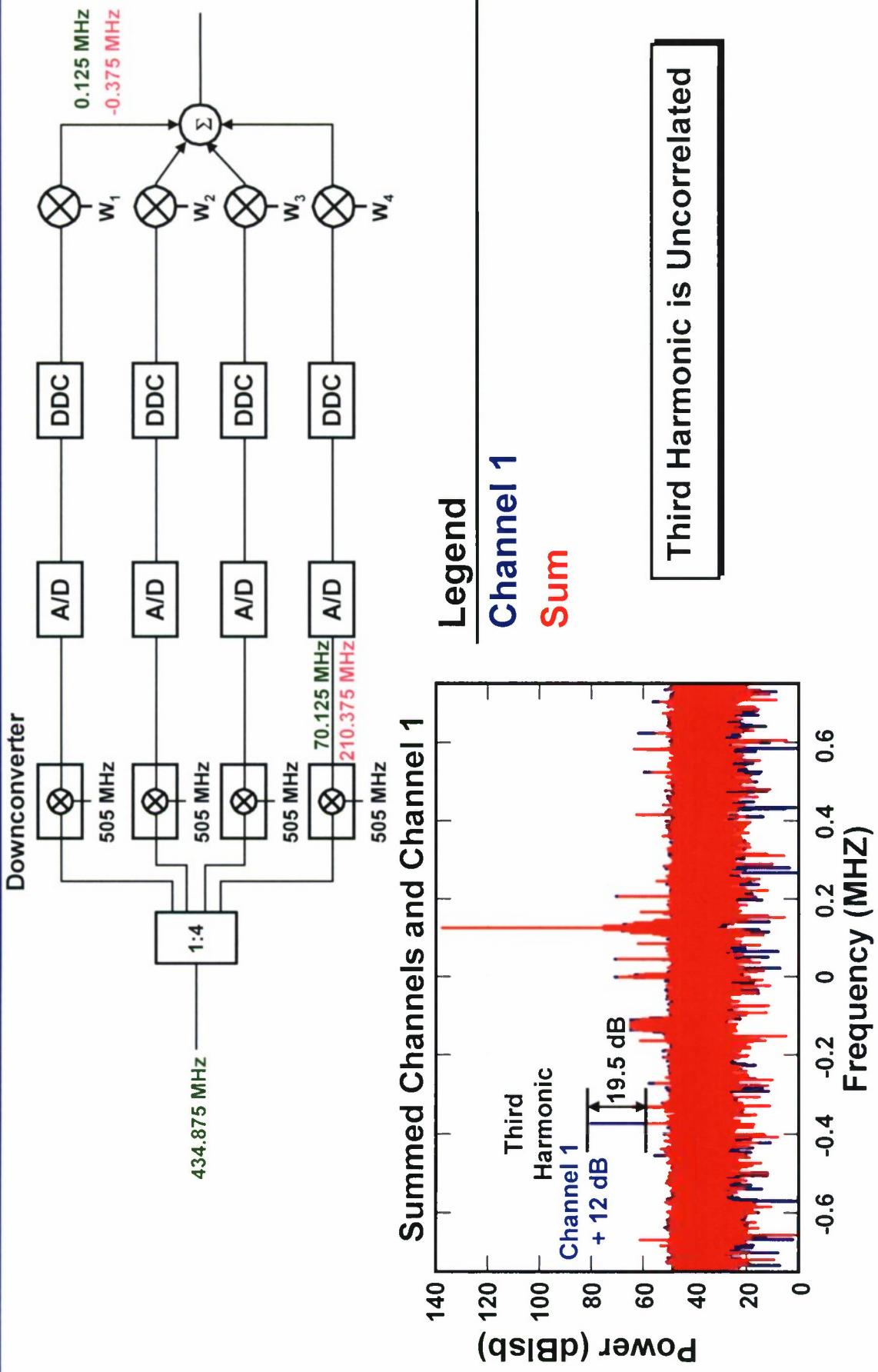


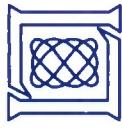
Third Harmonic Correlation





Third Harmonic Correlation





Outline

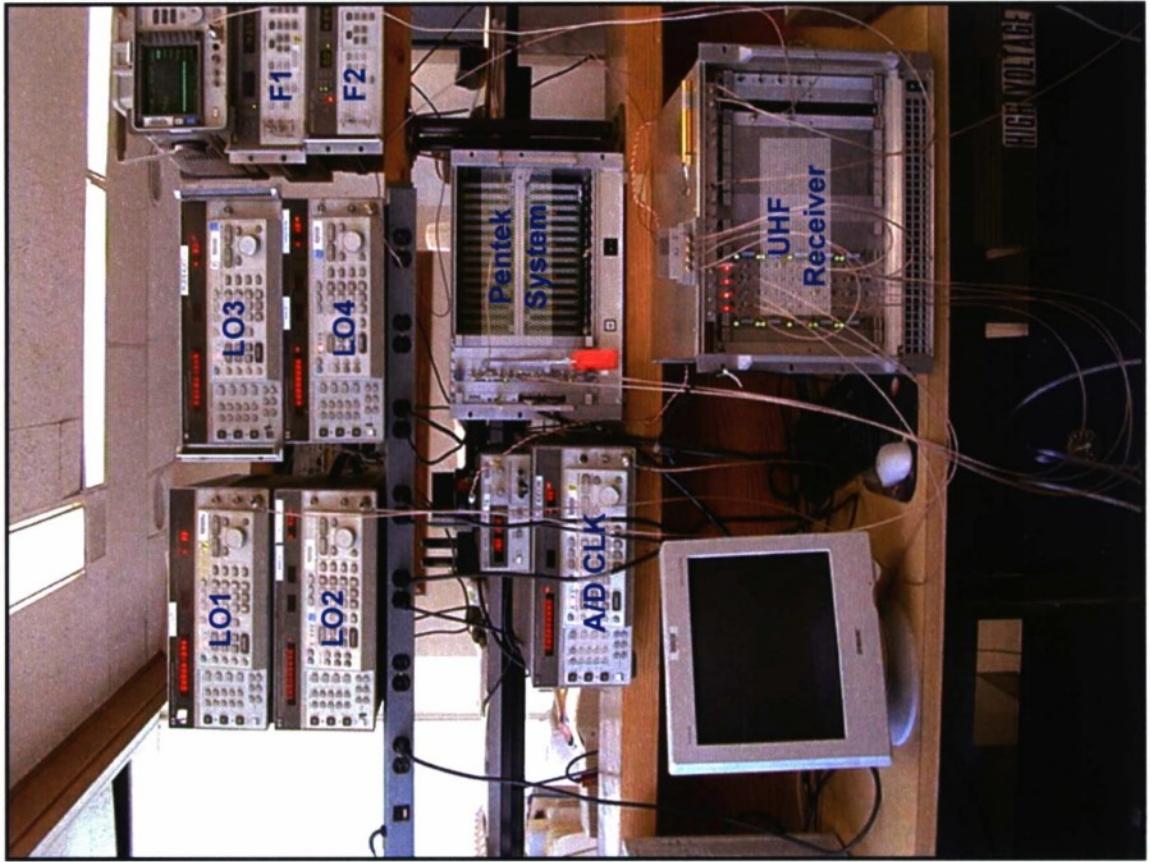
- **Introduction**
- **Importance of linearity and dynamic range**
- **Method for improving dynamic range**
- **Experimental data collection**
- **Future work and summary**

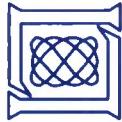




Future Work

- Shrink synthesizers
 - DDS
 - Phase-locked sources
- Other transformations
- Investigate more channels when Linear Array Upgrade hardware is available





Summary

- **Many sources of nonlinearities in a digital TR module**
- **Correlation of nonlinearities will reduce system dynamic range**
 - Measurements show that some distortions are strongly correlated, while others are not
- **Proposed method “Digital Array Nonlinearity Reduction” to make nonlinearities appear uncorrelated and/or incoherent from channel to channel**
 - Several manifestations described
 - Simulations used to examine the effect on some specific nonlinearities
 - Technique has been verified experimentally

REPORT DOCUMENTATION PAGE

Form Approved
OMB No. 0704-0188

Public reporting burden for this collection of information is estimated to average 1 hour per response, including the time for reviewing instructions, searching existing data sources, gathering and maintaining the data needed, and completing and reviewing this collection of information. Send comments regarding this burden estimate or any other aspect of this collection of information, including suggestions for reducing this burden to Department of Defense, Washington Headquarters Services, Directorate for Information Operations and Reports (0704-0188), 1215 Jefferson Davis Highway, Suite 1204, Arlington, VA 22202-4302. Respondents should be aware that notwithstanding any other provision of law, no person shall be subject to any penalty for failing to comply with a collection of information if it does not display a currently valid OMB control number. **PLEASE DO NOT RETURN YOUR FORM TO THE ABOVE ADDRESS.**

1. REPORT DATE (DD-MM-YYYY) 1 November 2004		2. REPORT TYPE		3. DATES COVERED (From - To)	
4. TITLE AND SUBTITLE		Correlation of Nonlinear Distortion in Digital Phased Arrays: Measurement and Mitigation		5a. CONTRACT NUMBER F19628-00-C-0002	
				5b. GRANT NUMBER	
				5c. PROGRAM ELEMENT NUMBER	
6. AUTHOR(S)		D. J. Rabideau and L. C. Howard		5d. PROJECT NUMBER	
				5e. TASK NUMBER	
				5f. WORK UNIT NUMBER	
7. PERFORMING ORGANIZATION NAME(S) AND ADDRESS(ES)		MIT Lincoln Laboratory 244 Wood Street Lexington, MA 02420-9108		8. PERFORMING ORGANIZATION REPORT NUMBER 101-1037	
9. SPONSORING / MONITORING AGENCY NAME(S) AND ADDRESS(ES)		Lincoln Laboratory Advanced Concepts Program 244 Wood Street Lexington, MA 02430-9108		10. SPONSOR/MONITOR'S ACRONYM(S)	
				11. SPONSOR/MONITOR'S REPORT NUMBER(S) ESC-TR-2004-086	
12. DISTRIBUTION / AVAILABILITY STATEMENT Approved for public release; distribution is unlimited.					
13. SUPPLEMENTARY NOTES					
14. ABSTRACT In a digital array, each receiver performs analog-to-digital-conversion (ADC), with the resulting digital data later combined via digital beamforming techniques. Since ADC is performed prior to beamforming, it is thus theoretically possible to enhance the dynamic-range (DR) of the individual receivers through post-ADC array integration gain. In practice, however, DR enhancement is limited by correlation of the nonlinearities (from receiver to receiver). Worse still, little published data exists on this subject (i.e., suitable for quantitatively assessing the correlation coefficients). This makes it difficult to predict how much DR enhancement will be achieved on real digital arrays. This presentation describes the results of recent experiments involving a four-channel digital receiver system. The system was used to measure the correlation (between receivers) of different types of nonlinear distortion. The measurements quantitatively demonstrate that some nonlinearities are highly correlated. Next, the system was used to evaluate a recently proposed method for decorrelating nonlinear distortion in digital arrays. The measurements show that the mitigation technique is successful in decorrelating some nonlinear signal components.					
15. SUBJECT TERMS arrays radar phase noise intermodulation digital arrays dynamic range spurs digital receivers					
16. SECURITY CLASSIFICATION OF:			17. LIMITATION OF ABSTRACT	18. NUMBER OF PAGES	19a. NAME OF RESPONSIBLE PERSON
a. REPORT Unclassified	b. ABSTRACT Unclassified	c. THIS PAGE Unclassified	Same as report	43	19b. TELEPHONE NUMBER (include area code)